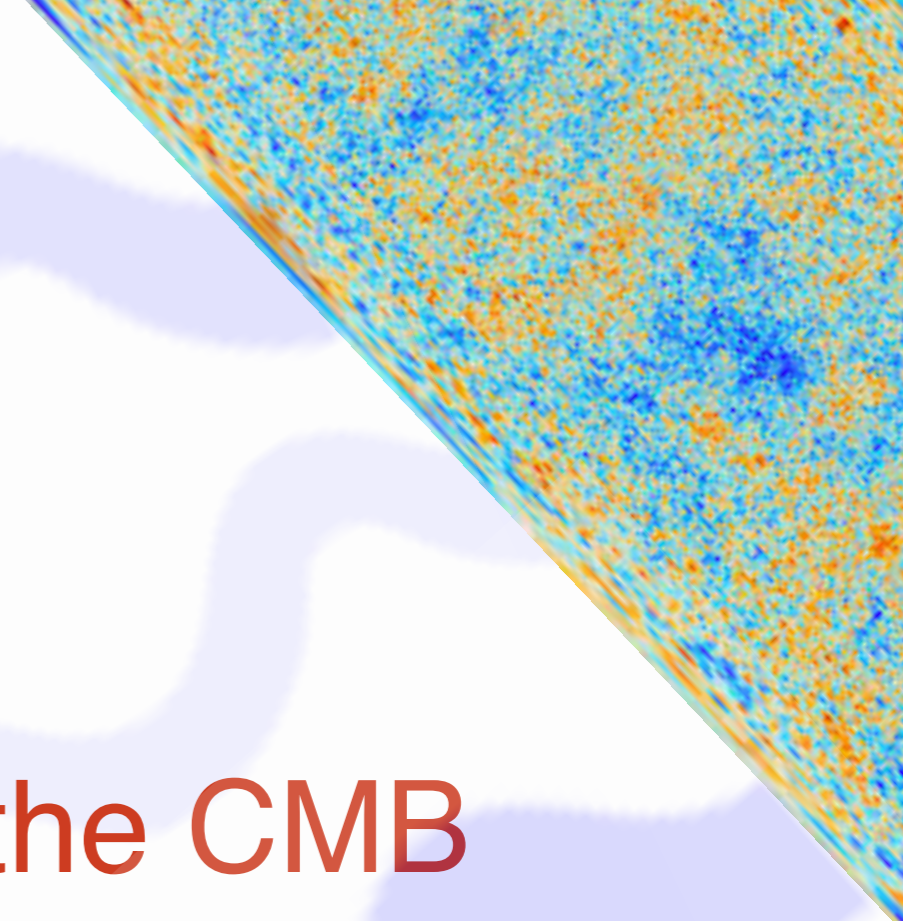
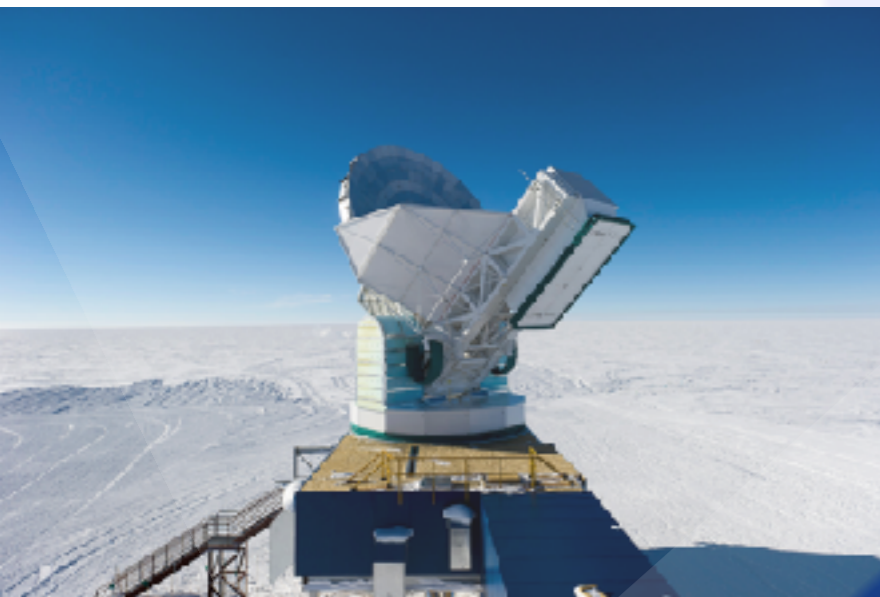


# How we really measure the CMB

Ritoban Basu Thakur  
Astro 448

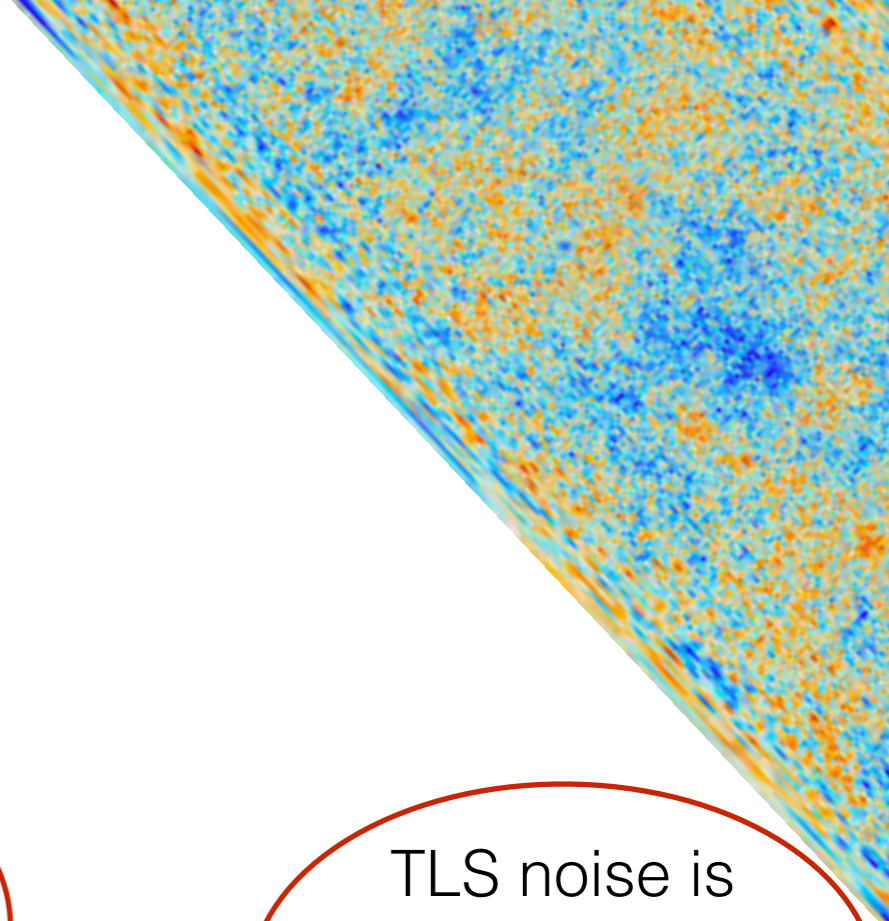


Did I plug in the amplifier ?

Isocurvature fluctuations and the slope of slow-roll...

Did I include that minus sign ?

TLS noise is subdominant compared to SQUID ...



The what-to and how-to of measuring ?

How to get the light in ?

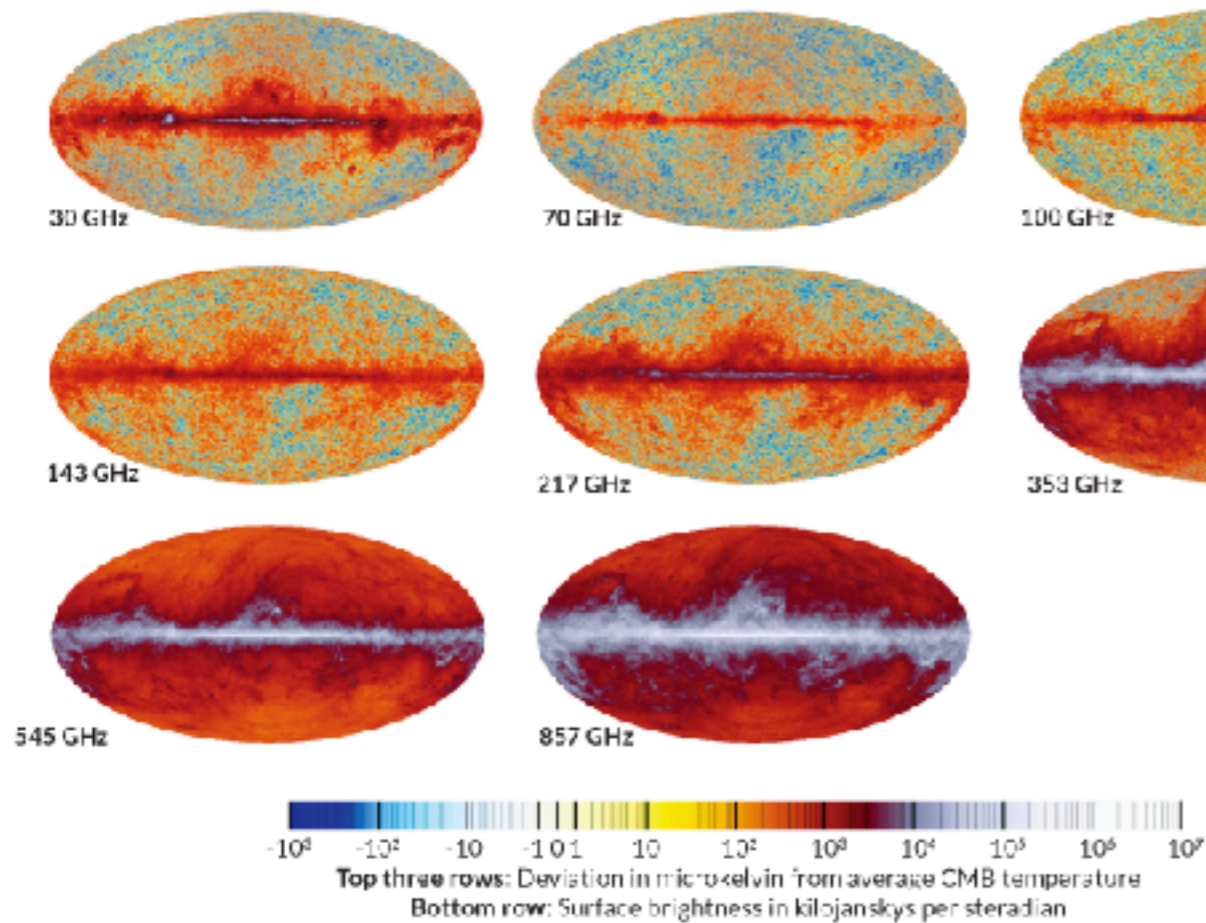
How to measure the light ?

Some detectors in detail

**The light**

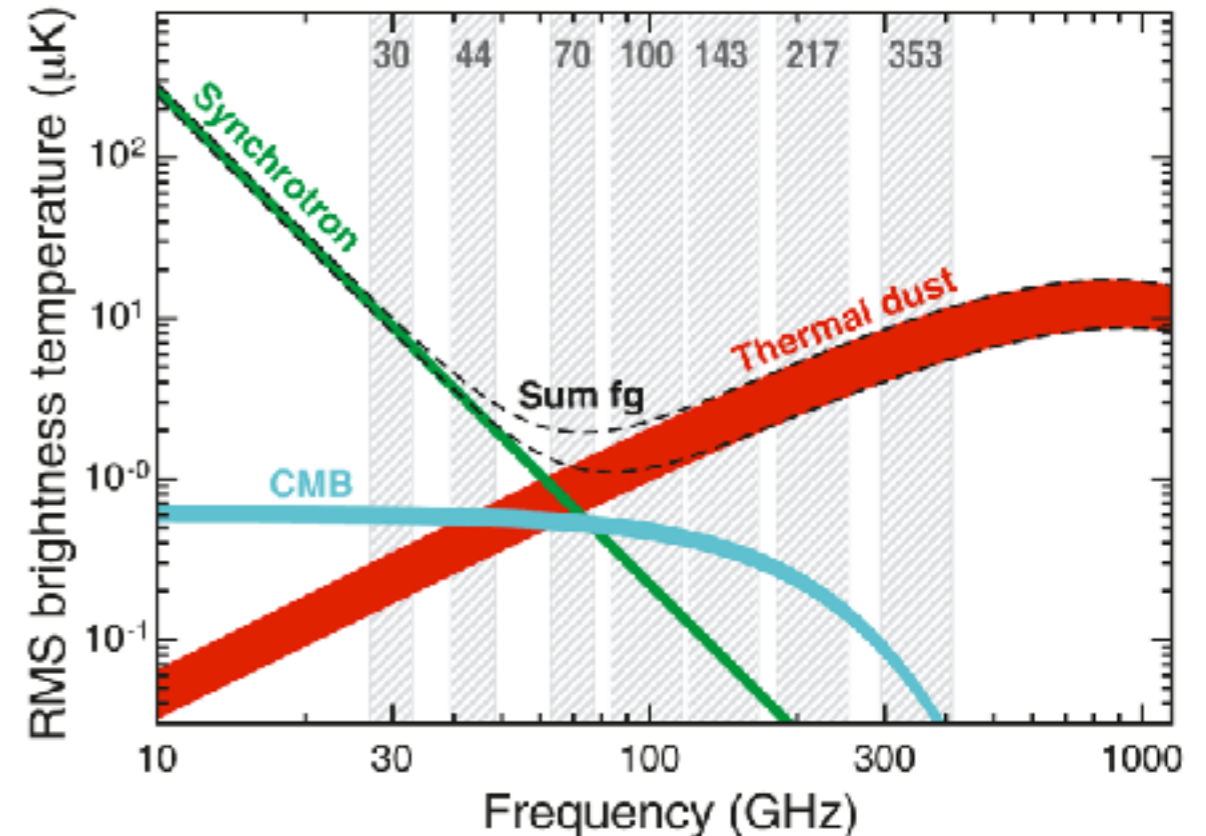
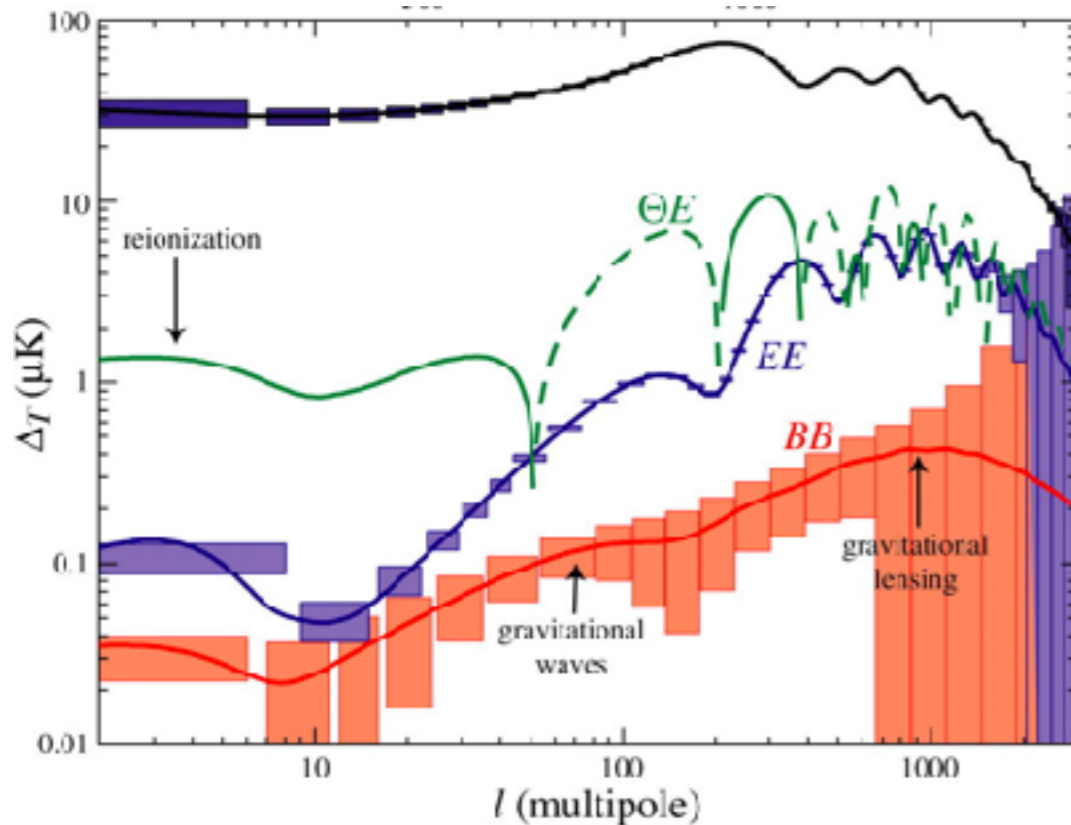


# The light

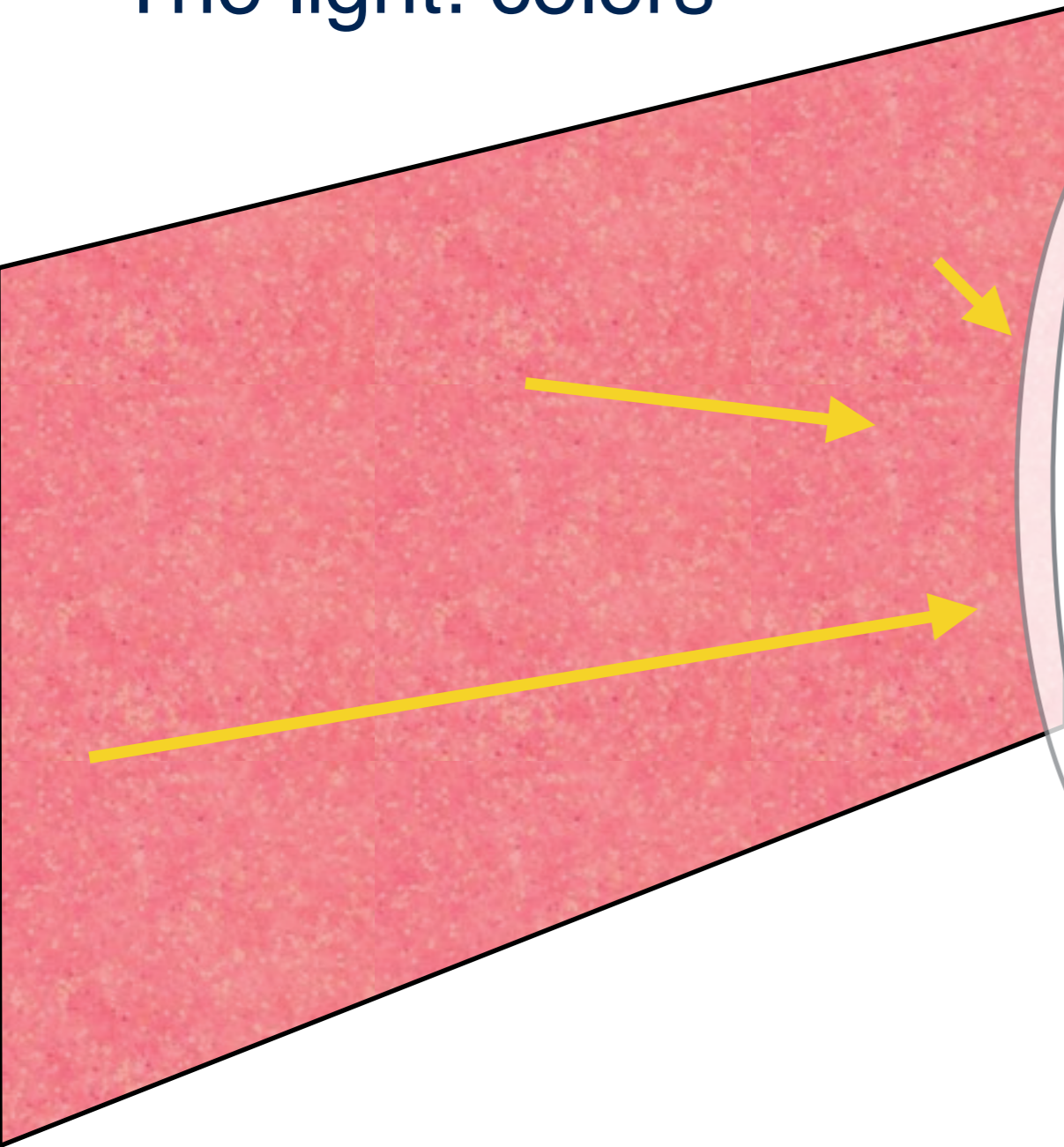


CMB is  $\sim 2.7$  K Black Body, with maximal signal over foregrounds  $\sim 100$  GHz

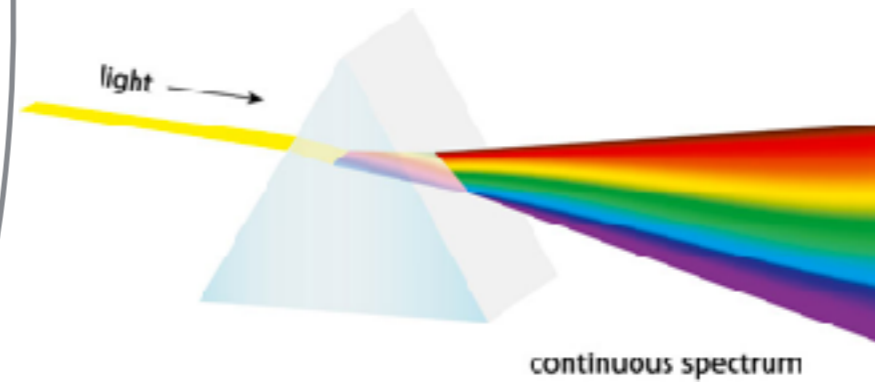
Tiny intensity fluctuations in frequency and location contain a plethora of interesting physics



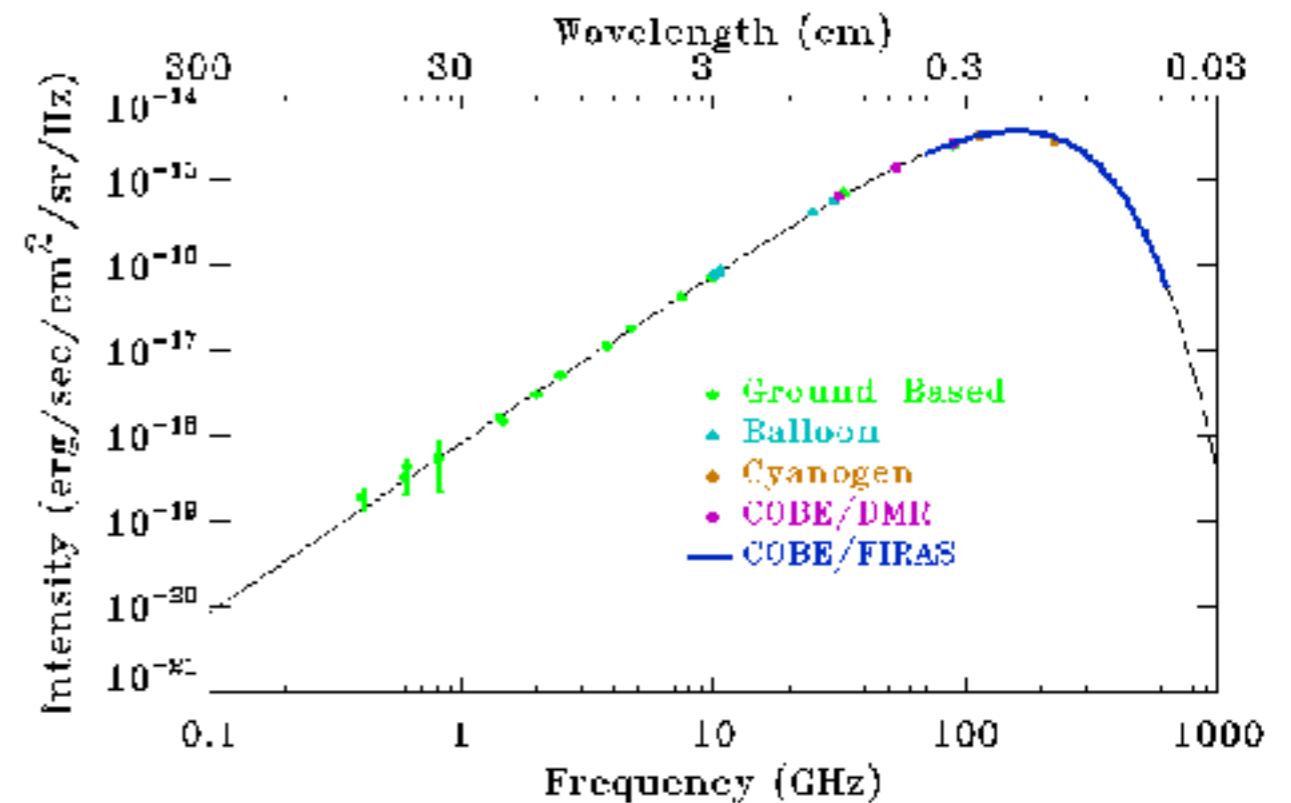
# The light: colors



Large area collector  
+ collimator



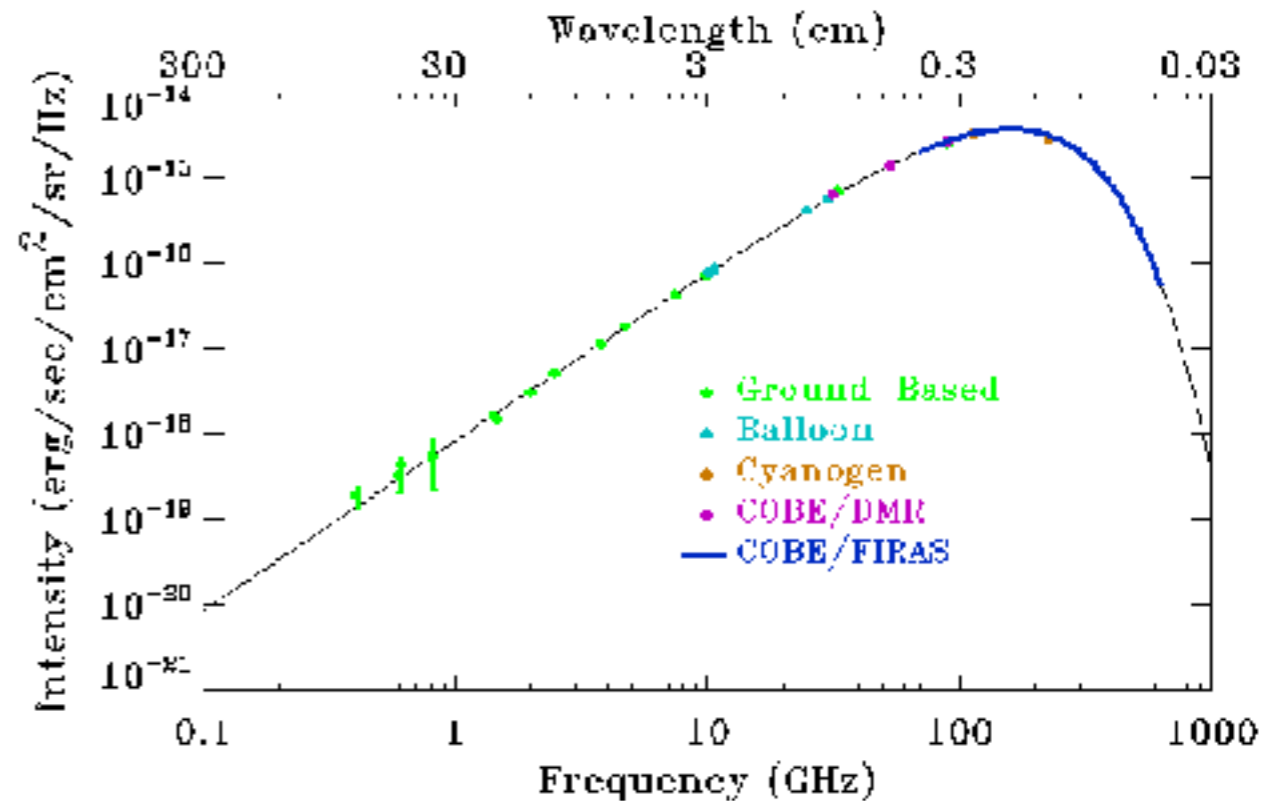
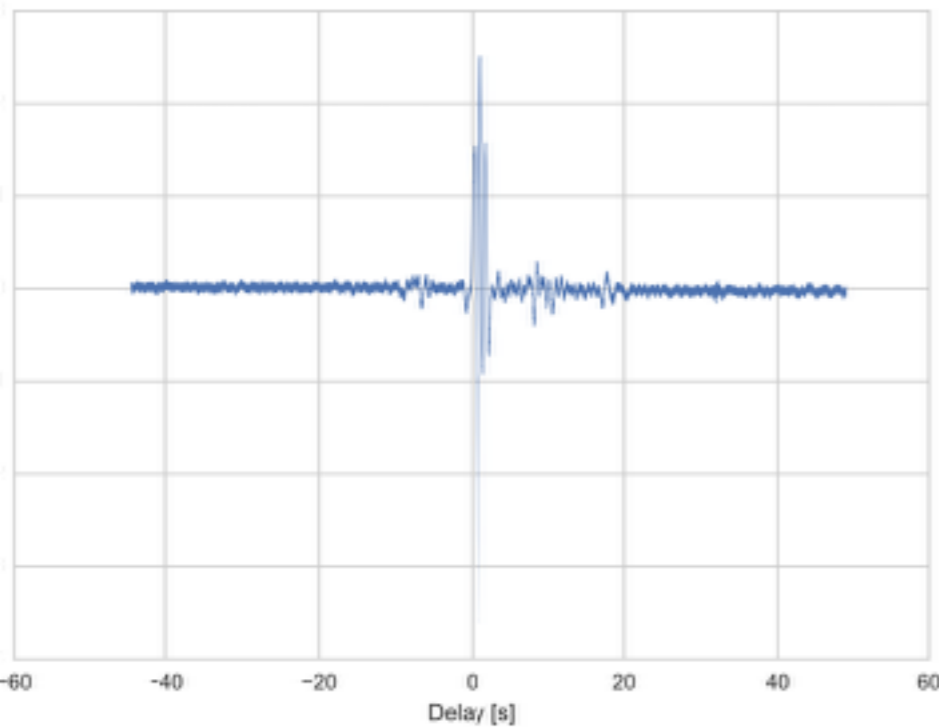
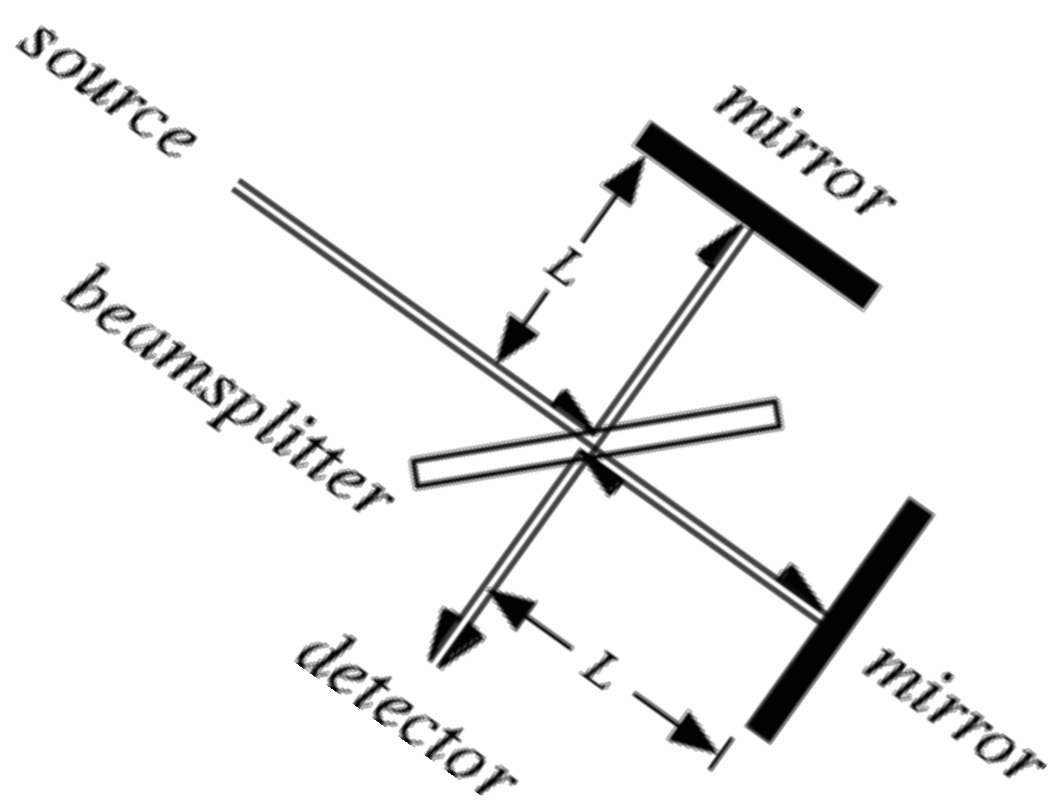
Broad-band 2.7 K BB  
spectrum measured  
across three decades



# The light: colors

Not prisms ...

Fourier Transform Spectrometry  
at the heart measurements.

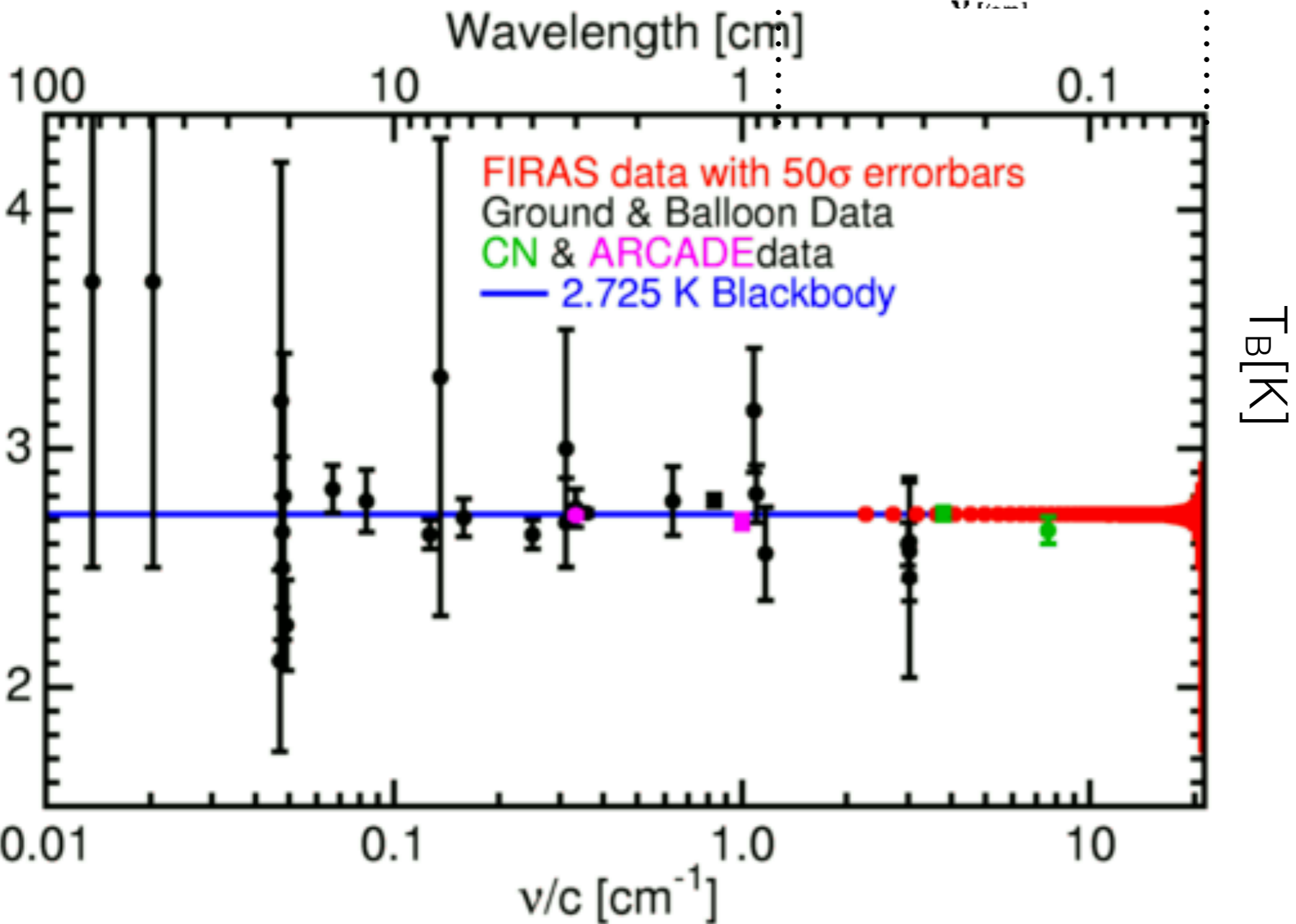
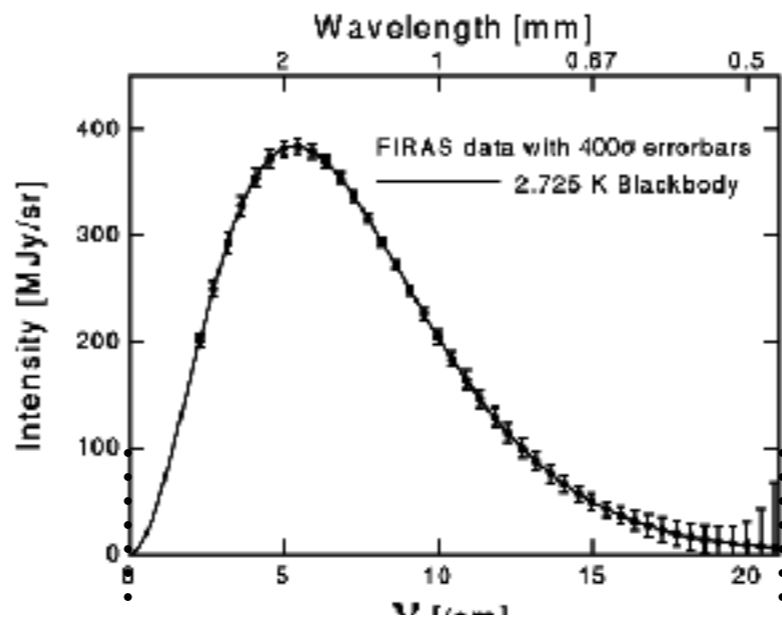


$$\tilde{I}(\lambda^{-1}) = 4 \int_{-L}^L \{I(\Delta) - 0.5 I(\Delta = 0)\} \cos(2\pi\nu\Delta/\lambda)$$

Fourier transform of Interferogram



# The light: colors

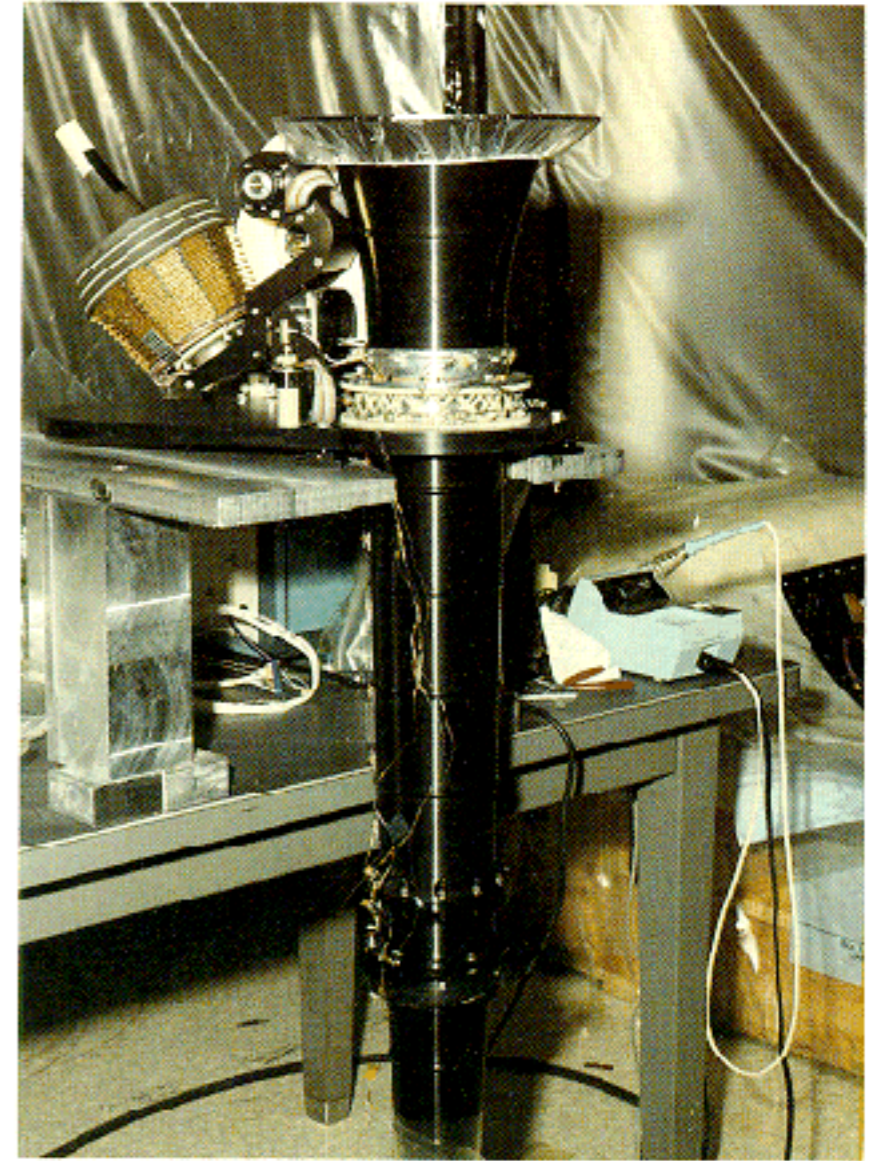
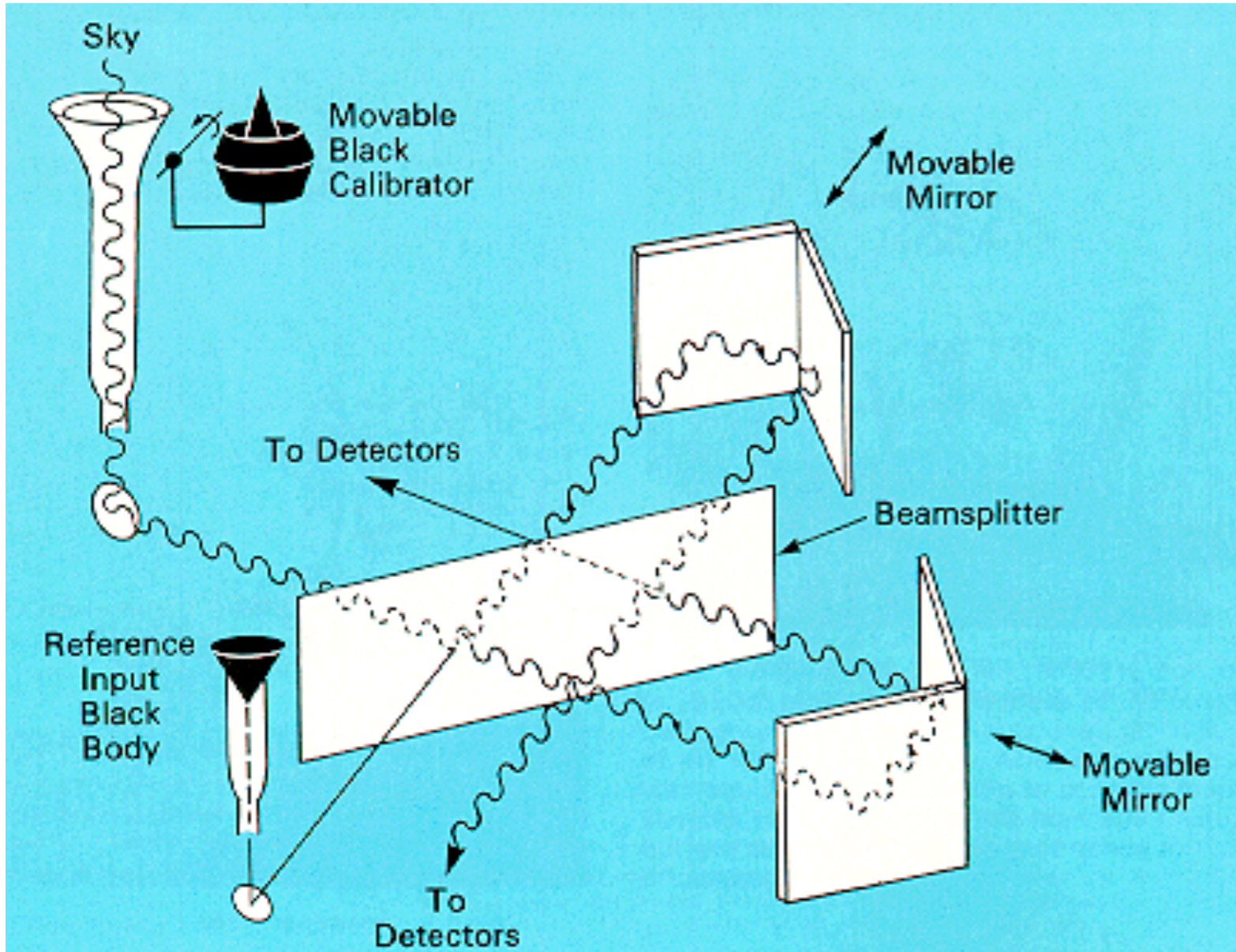


COBE FIRAS  
has the best  
spectral  
measurements,  
constraining  
 $dI/I$  to  $\lesssim 10^{-5}$



# The light: colors

## COBE Far Infrared Absolute Spectrophotometer (FIRAS)



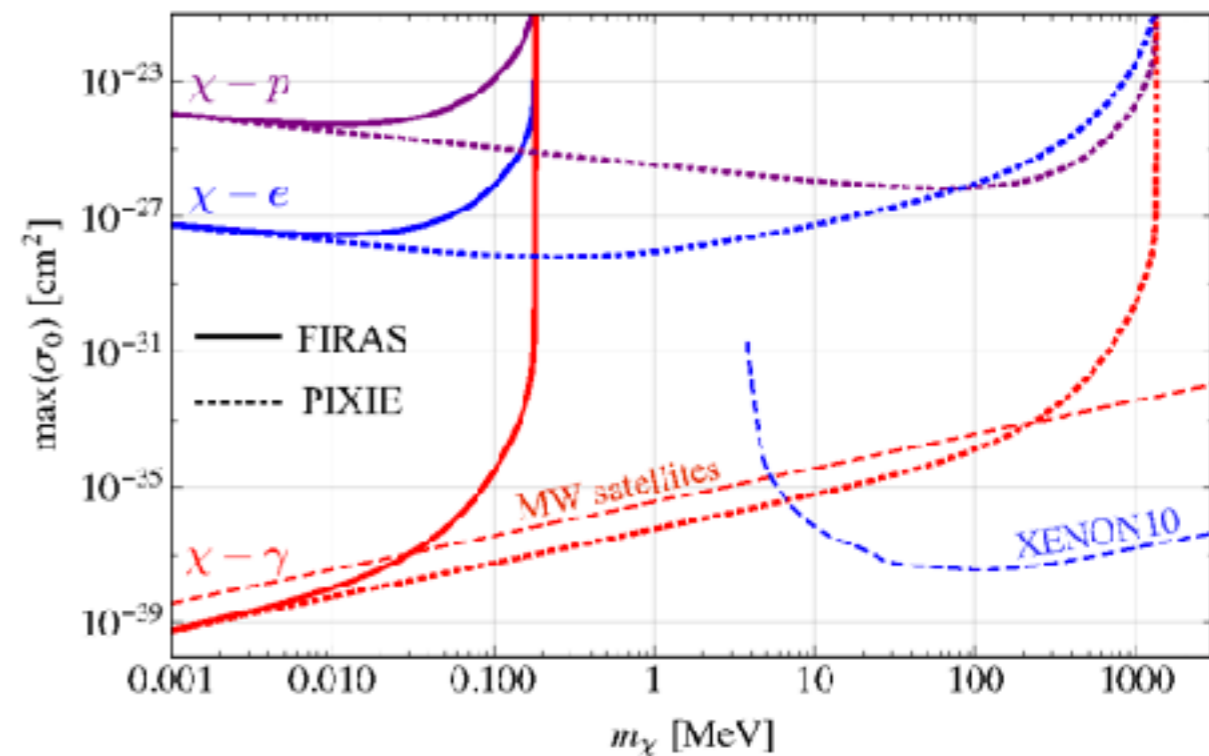
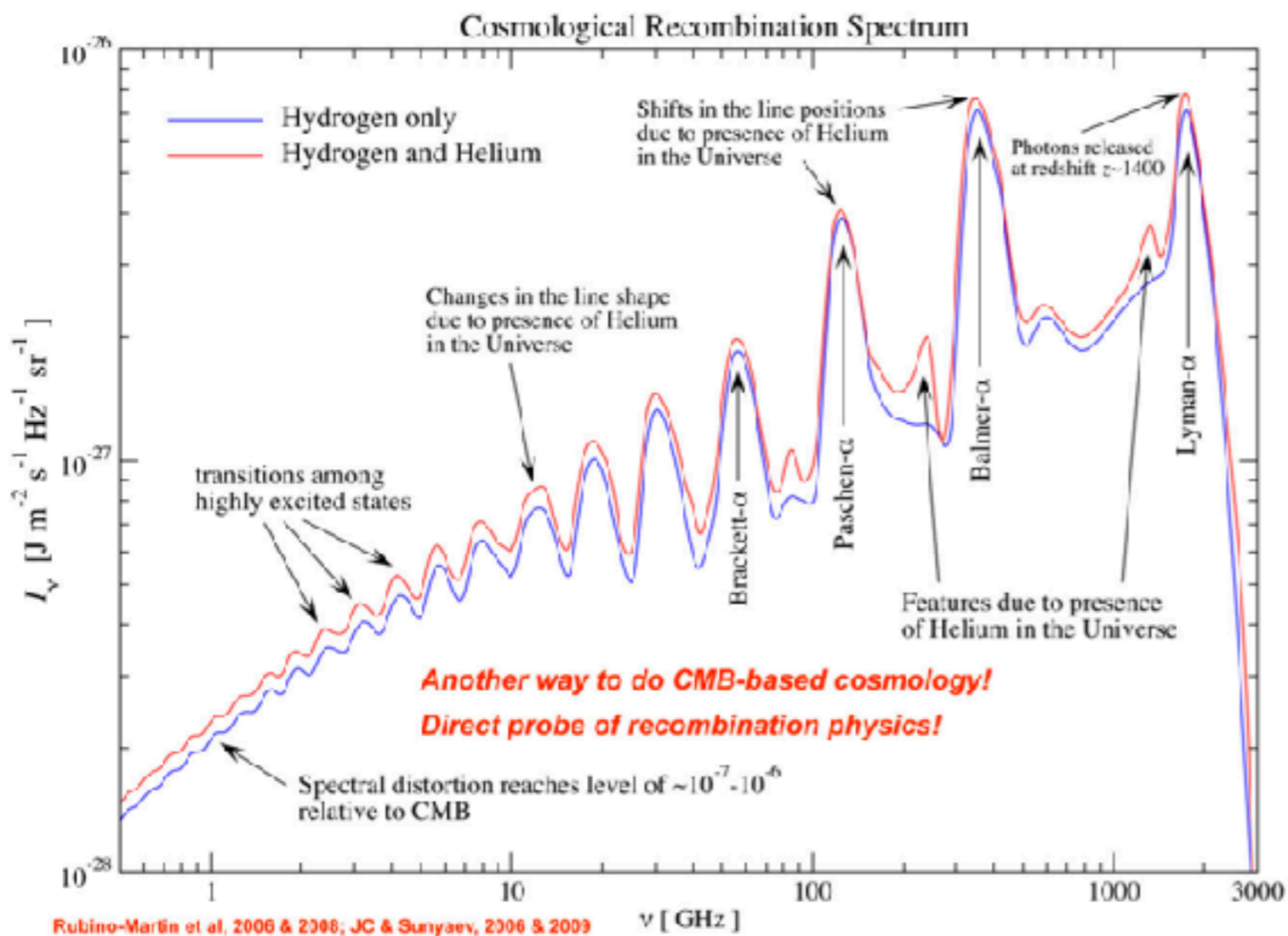
*Horn antenna with movable calibrator. Protective plastic covers will be removed.*

Fits to the cosmic spectral distortion parameters give 95% confidence limits of  $|\mu/kT| < 3.3 \times 10^{-4}$  and  $|y| < 2.5 \times 10^{-5}$ .



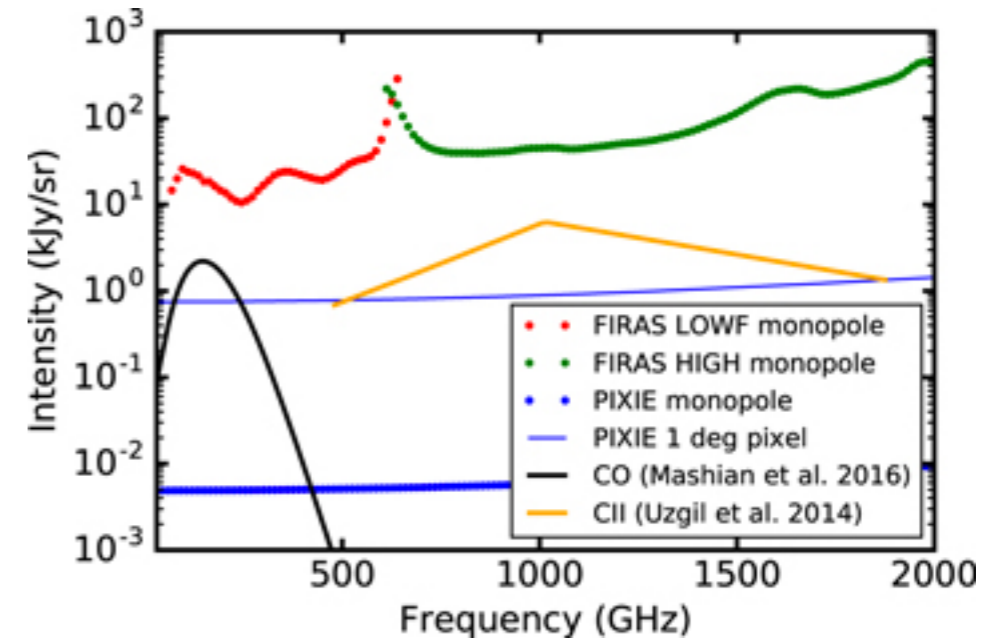
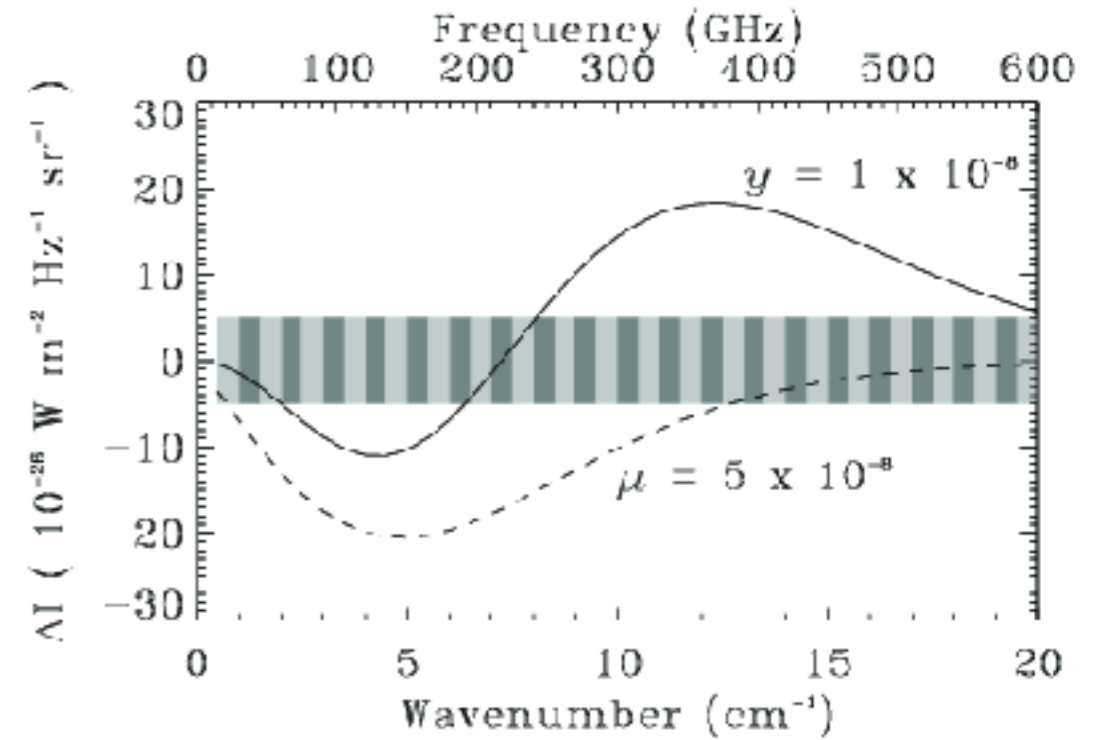
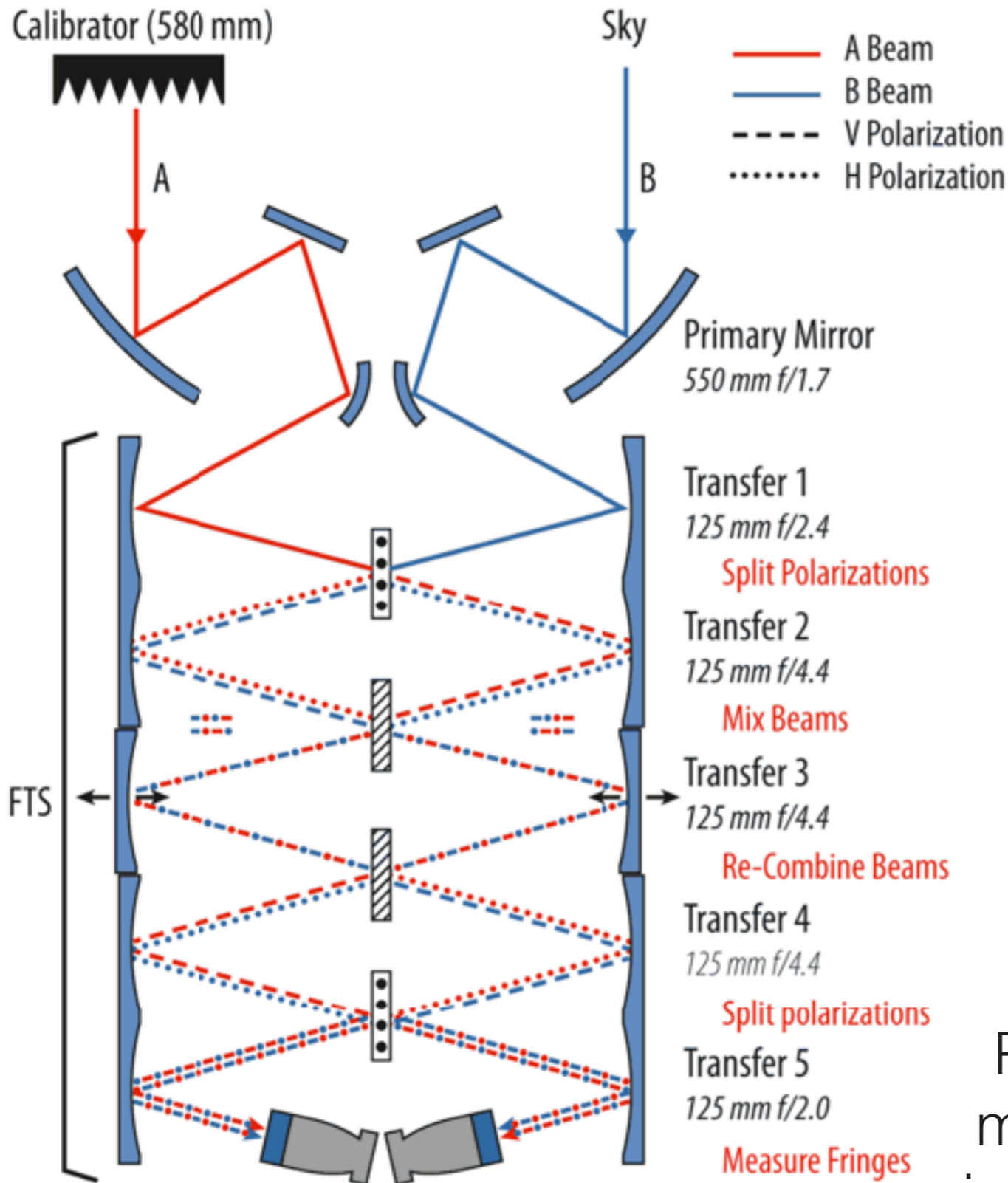
# The light: colors

Lot of interesting physics in CMB Spectral Distortions



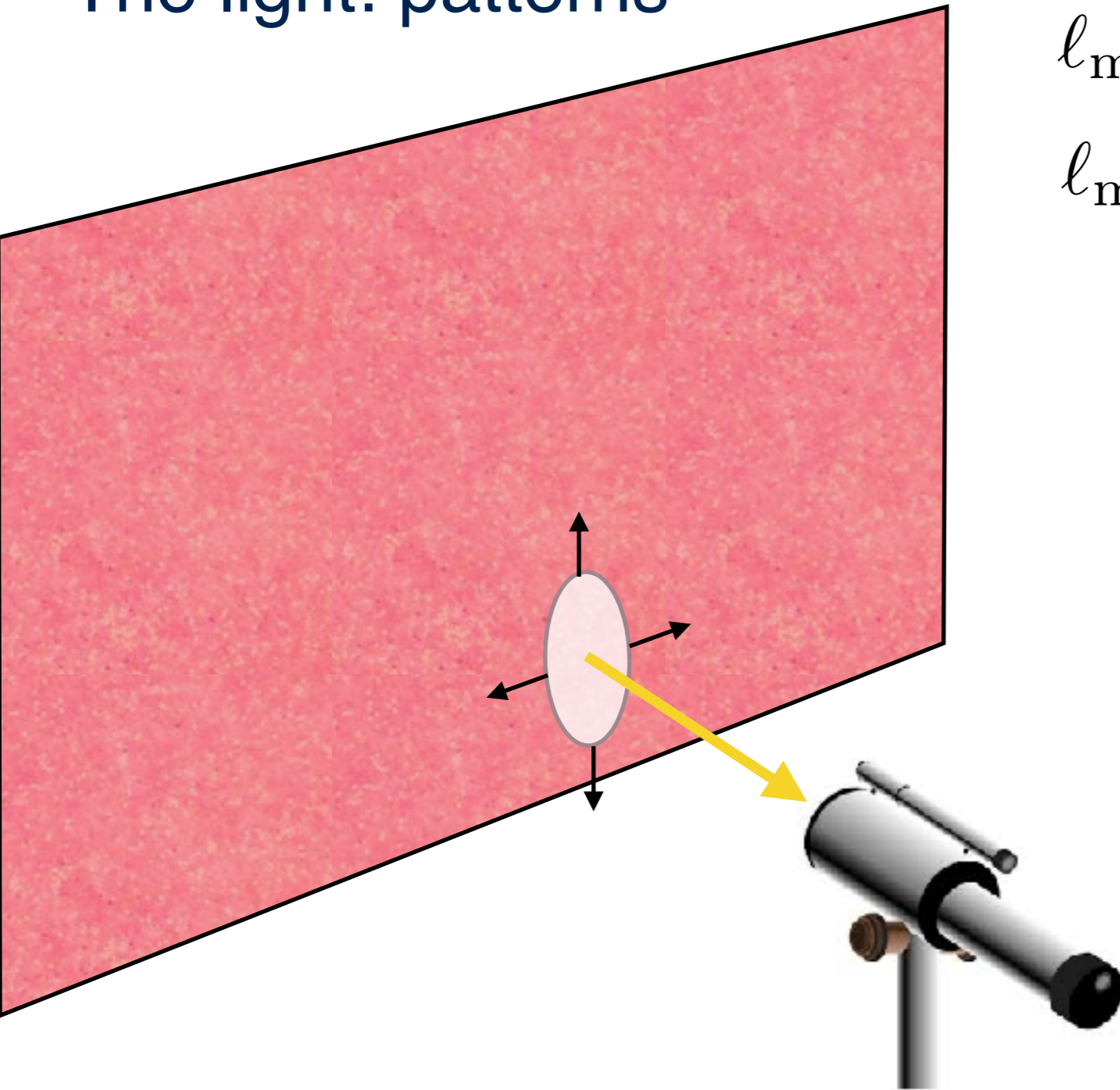
Probing energy injection from DM interactions / decays  
Recombination lines etc. Several papers by Jens Chulba et al.

# The light: colors



Proposed PIXIE satellite FTS can do much better than FIRAS can open up interesting avenues. Ask Steve Meyer!

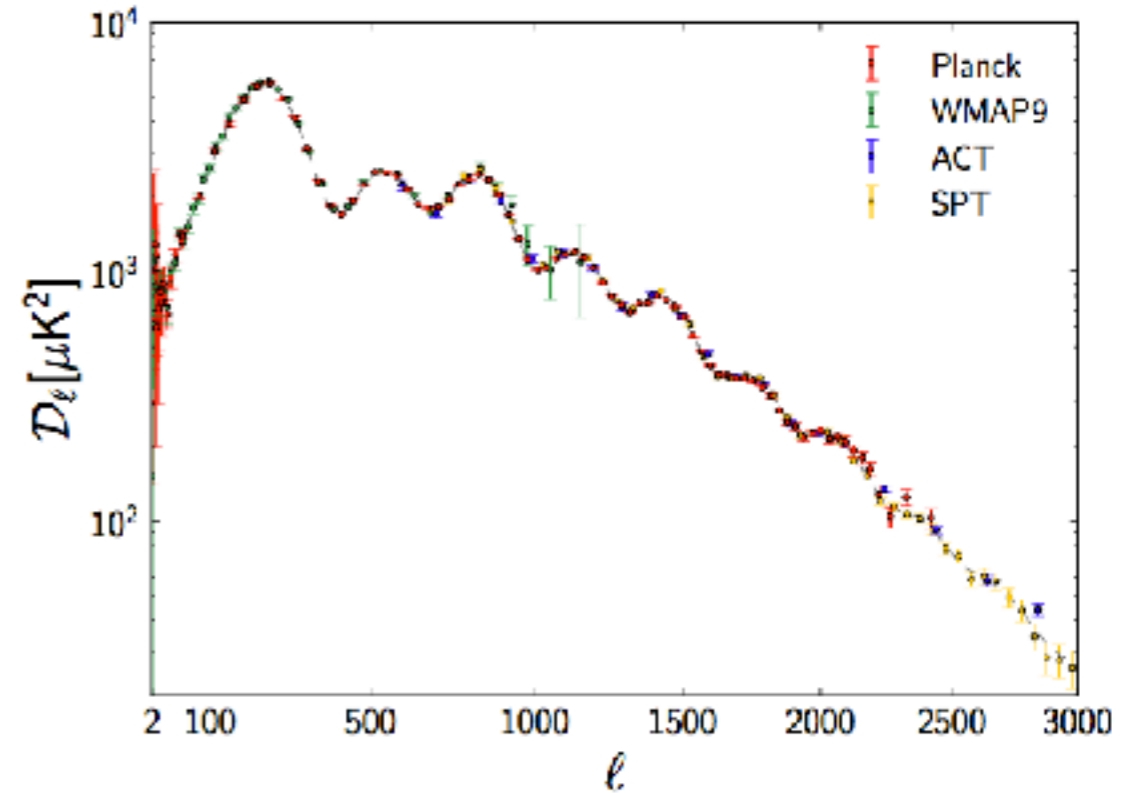
# The light: patterns



Small(ish) area collector  
+ Scanning

$$\ell_{\min} = \pi / (N \theta_{\text{res}}) \quad \text{Map-size}$$

$$\ell_{\max} = \pi / \theta_{\text{res}} \quad \text{Beam-size}$$



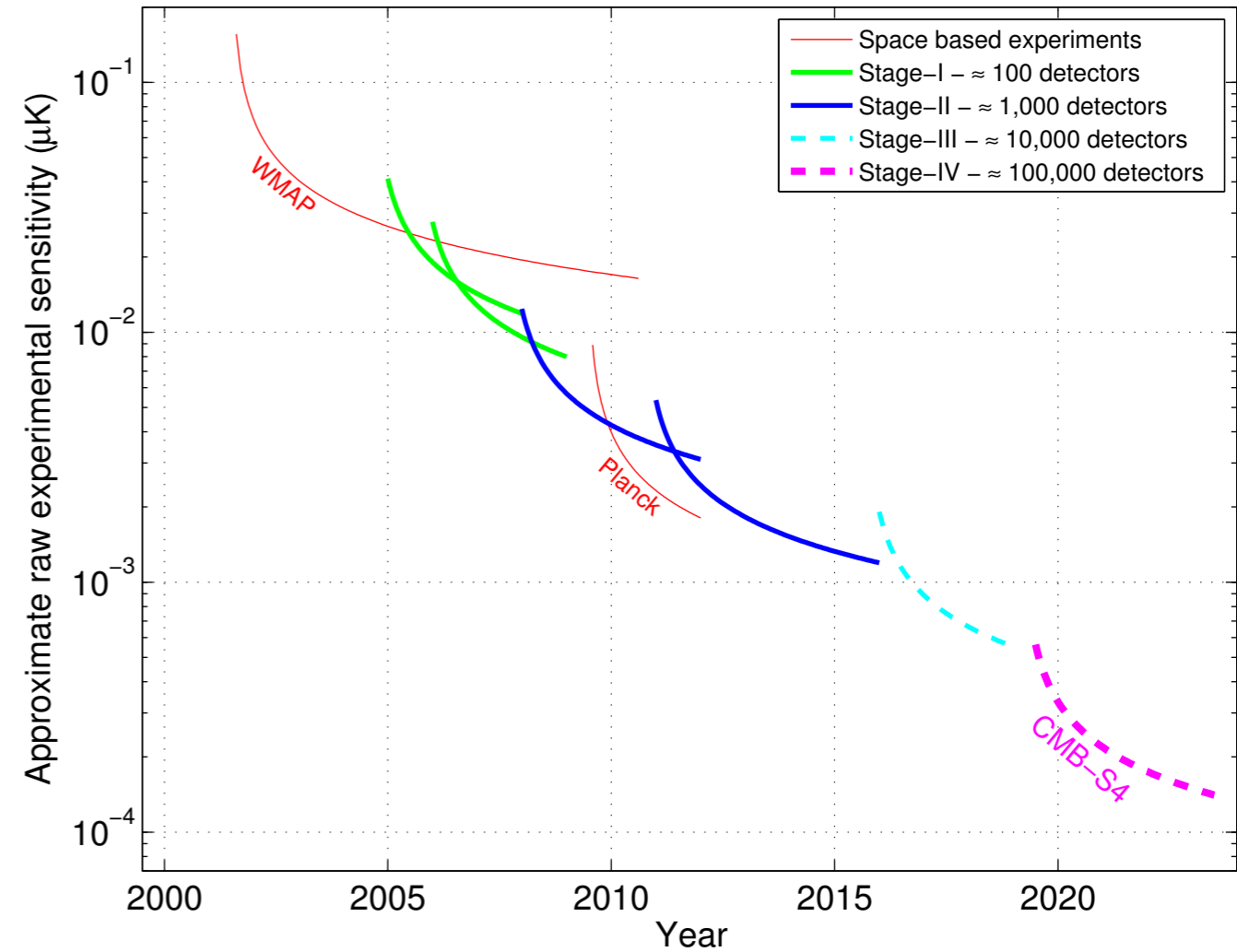
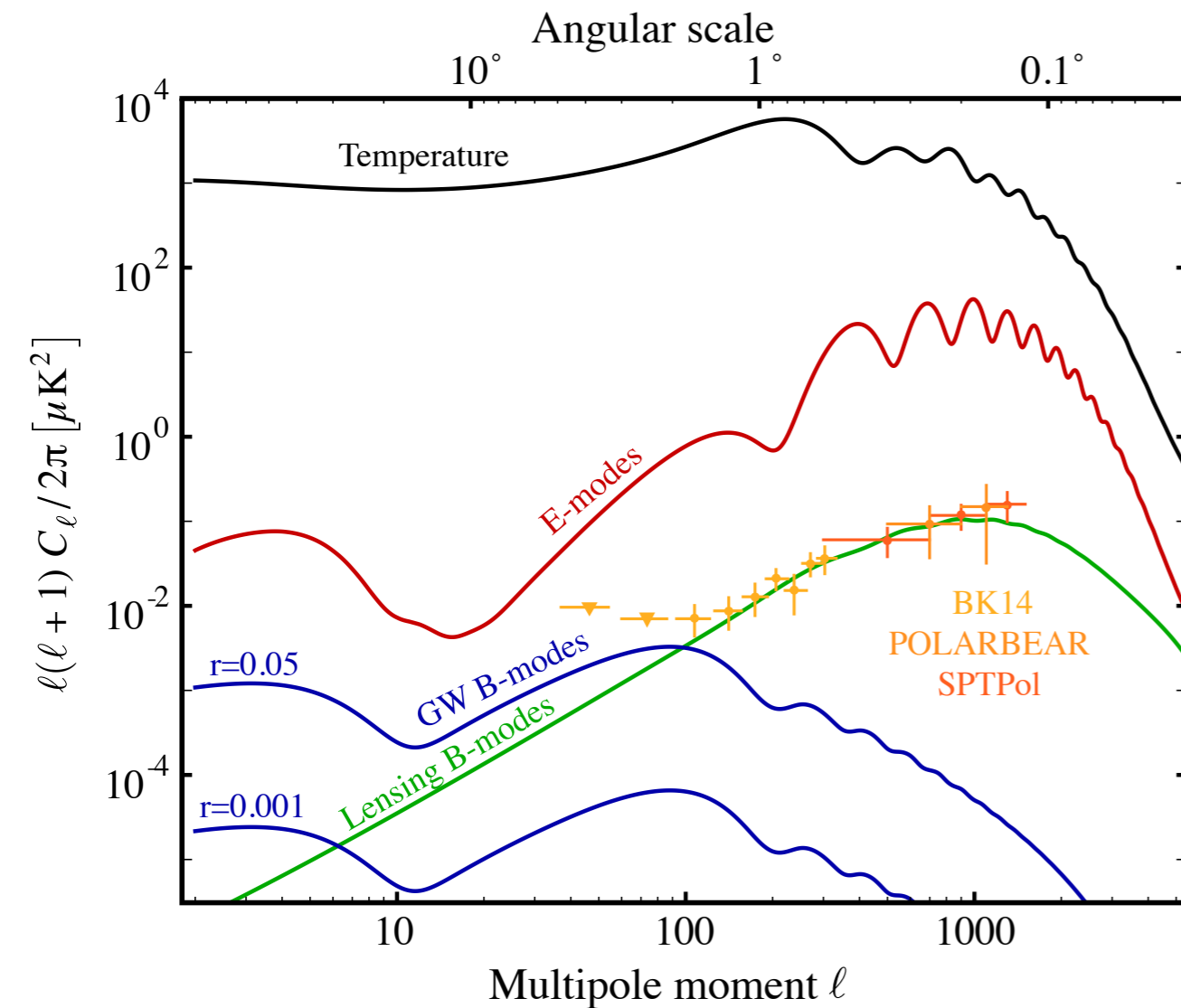
Relation between time/frequency  
and ell, given a scan speed

$$\frac{f [\text{Hz}]}{\ell} = \frac{v_x [\text{rad/s}]}{\pi} = \frac{v_x [\text{deg/s}]}{180}$$



# The light: patterns

Very roughly, sensitivity to some  $\Delta T$  scales with # detectors



Noise-Equivalent-Temperature (NET) of a detector is measured /estimated.

Temperature-power of desired CMB feature is noted ( $P_x$ )

$$\text{NET}/N_{\text{dets}} < P_x$$

In detail ...

# The light: patterns

Suppose we want to *discover* E-modes at  $\ell \sim 10^3$

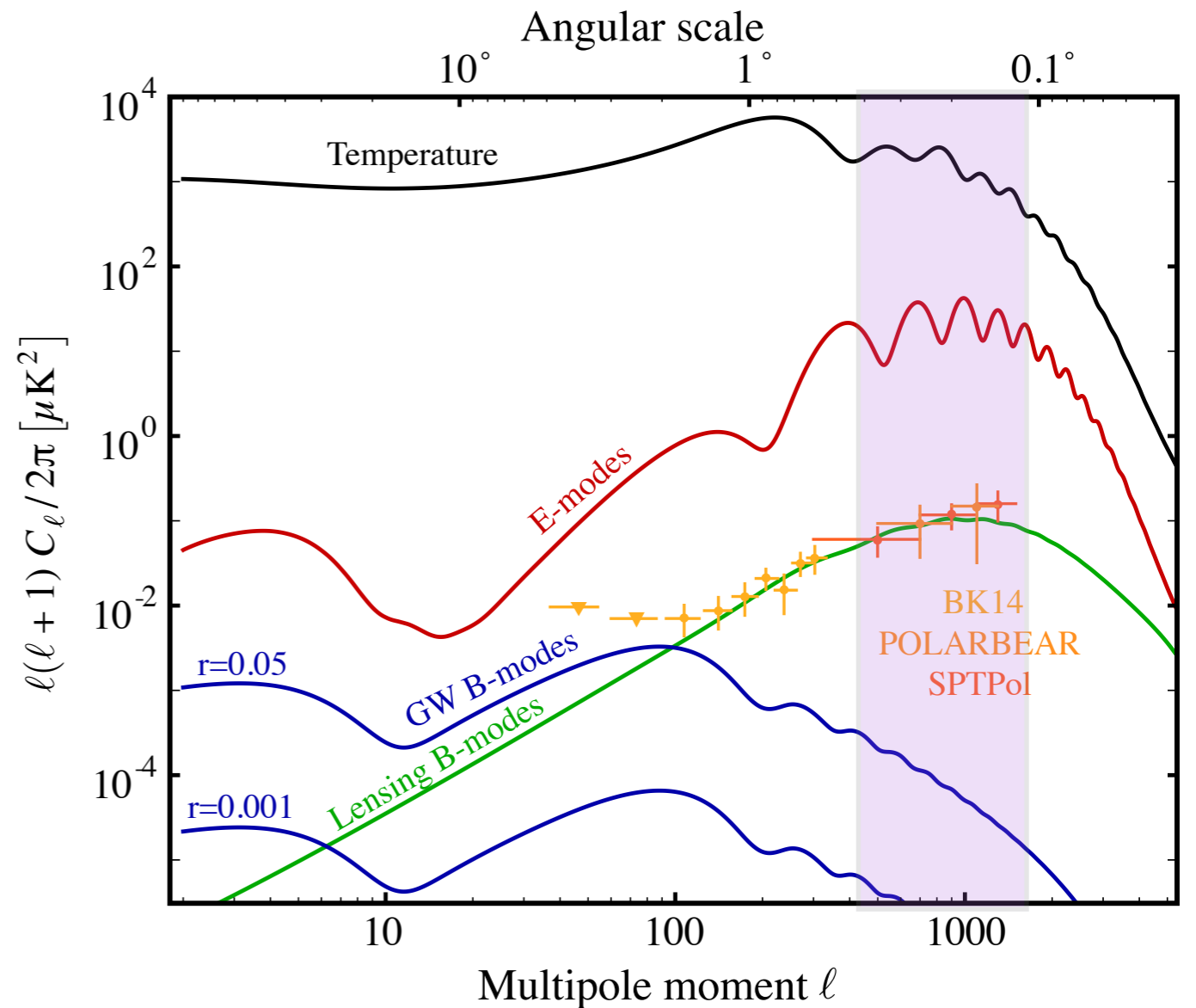
Beam has to small enough  
Need to resolve our mode

$$\theta_{\text{res}} [\text{rad}] < \pi / (\ell = 10^3)$$

Sky has to be big enough  
Need to capture a wavelength

$$\theta_{\text{sky}} [\text{rad}] > \pi / (\ell = 10^3)$$

Here sky area mapped is  $\sim \theta_{\text{sky}}^2$



\*for a real experiment this will be  $> 100$  Hz

# The light: patterns

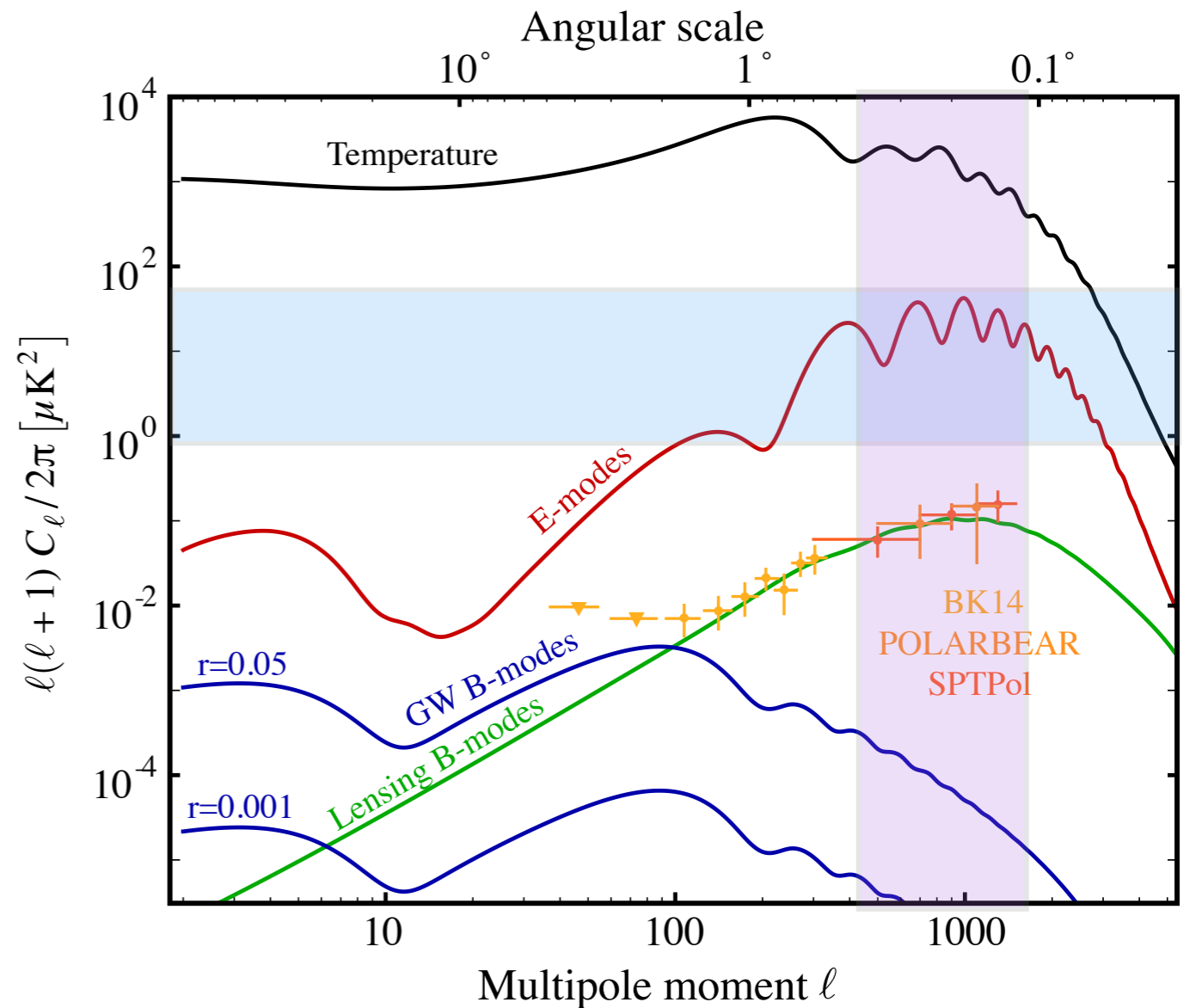
Suppose we want to *discover* E-modes at  $\ell \sim 10^3$

The experiment needs sensitivity  $\mathcal{D}_\ell > 1 \mu\text{K}^2$  or  $\mathcal{C}_\ell > 2\pi (\text{nK rad})^2$

Suppose that we are scanning at  $1 \text{ deg/sec} = \pi/180 \text{ rad/s}$

->  $f(\ell = 10^3) = 10^3/180 = 5.55 \text{ Hz}$

-> Sampling rate  $\gg 11.11 \text{ Hz}^*$   
and sampling time  $\gg 0.18 \text{ sec}$



For every 0.18 sec of scanning time we collect one more  $\ell = 10^3$  mode

**The light: patterns**     Suppose we want to *discover* E-modes at  $\ell \sim 10^3$

The experiment needs sensitivity  $e_\ell > 2\pi$  (nK rad)<sup>2</sup>

Suppose that *one* detector has noise RMS given by  $w_1$  [ $\mu\text{K}/\sqrt{\text{Hz}}$ ], simplified white noise power spectral density

Suppose we have  $N_{\text{dets}}$  and we are scanning for  $T_s$  ( $\gg 0.18$  s) s

We know that the noise variance will scale with  $1/N_{\text{dets}}$

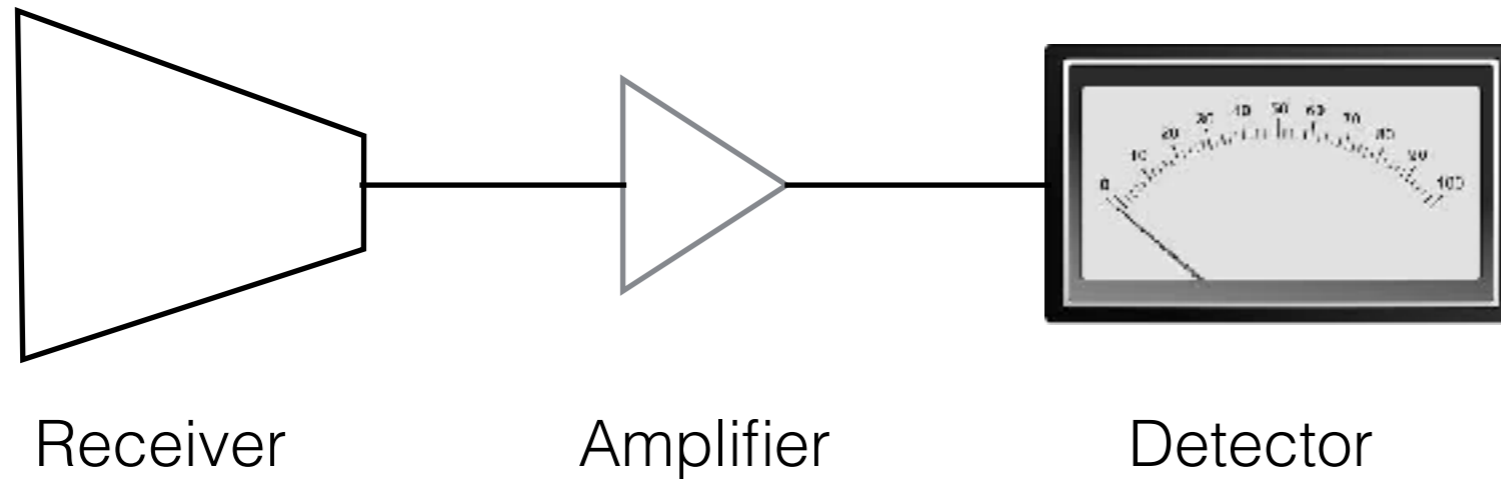
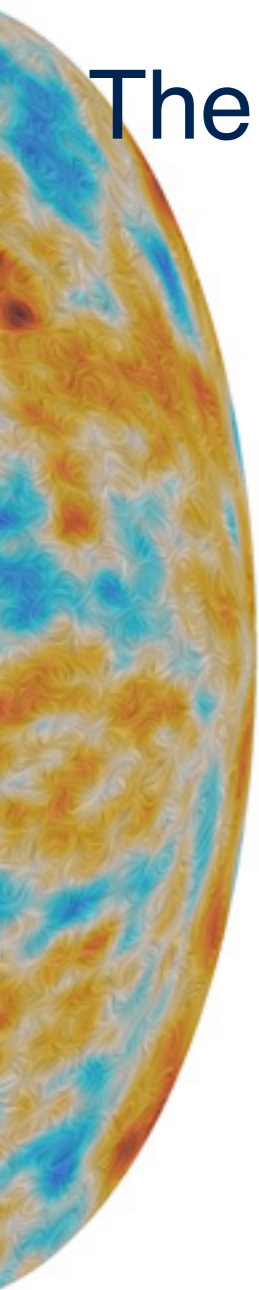
Longer scan duration ( $T_s$ )  $\rightarrow$  more modes captured, i.e. higher SNR

$$\frac{w_1^2 \theta_{\text{sky}}^2}{N_{\text{dets}} T_s} \leq 2\pi [\text{nK-rad}]^2$$

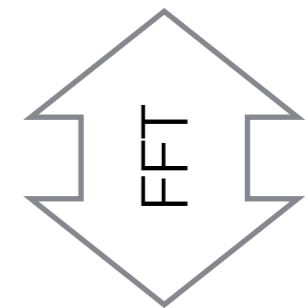
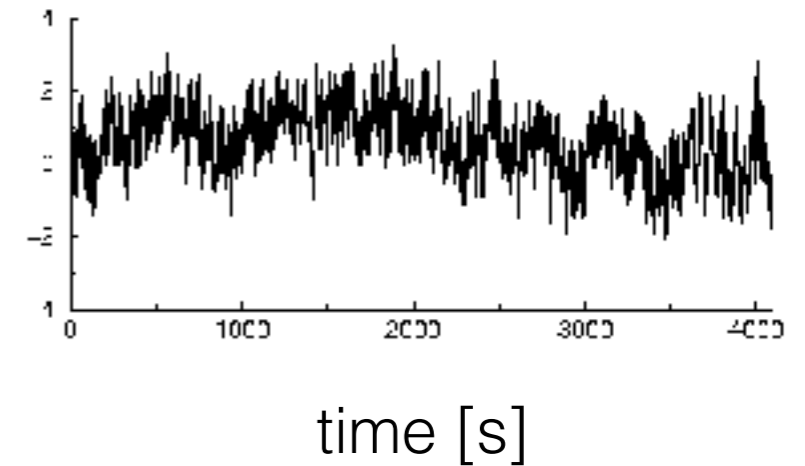
This is the (Knox) sensitivity limit for our case



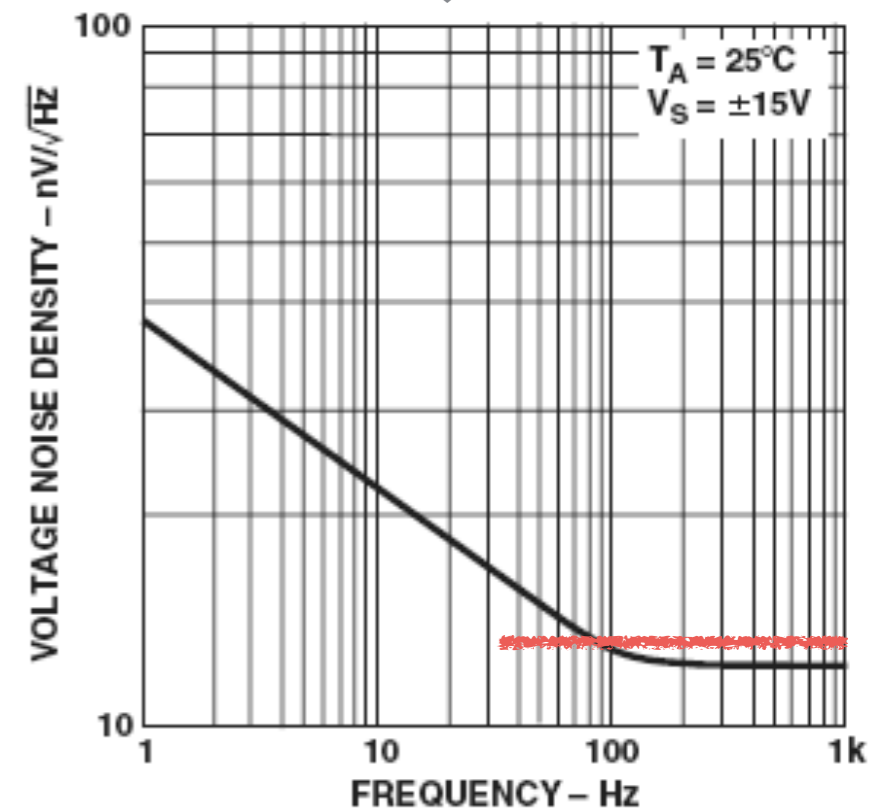
# The light: patterns



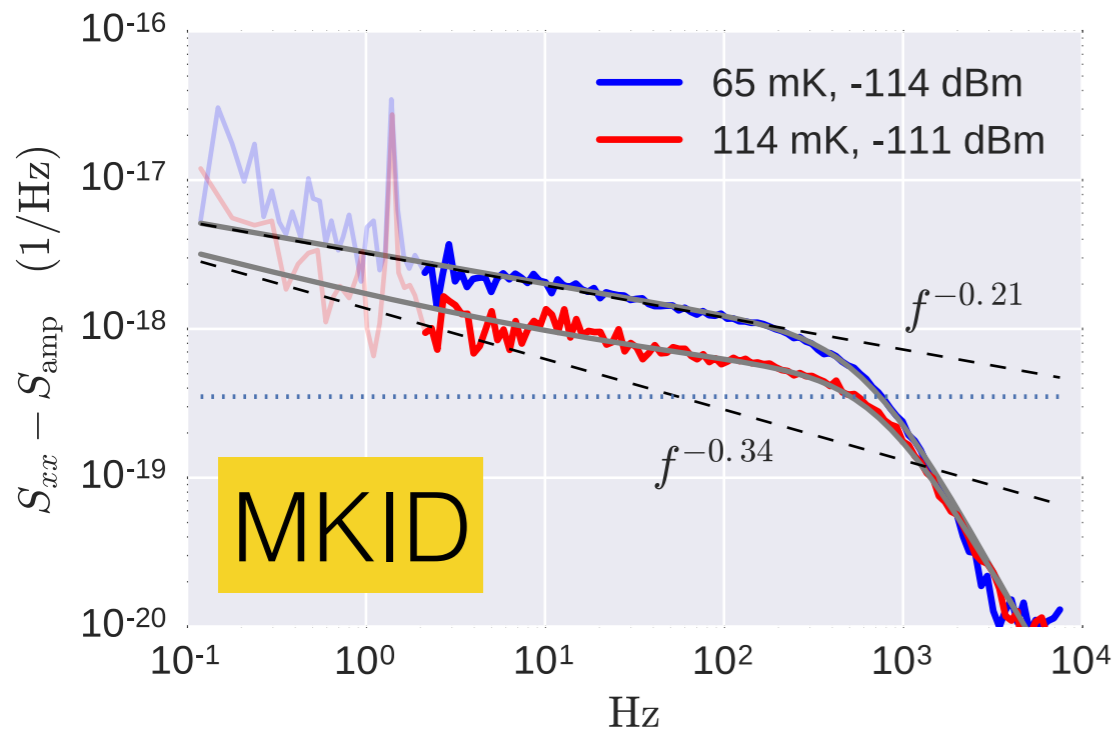
detector output [nV, Hz,  $\mu$ A...]



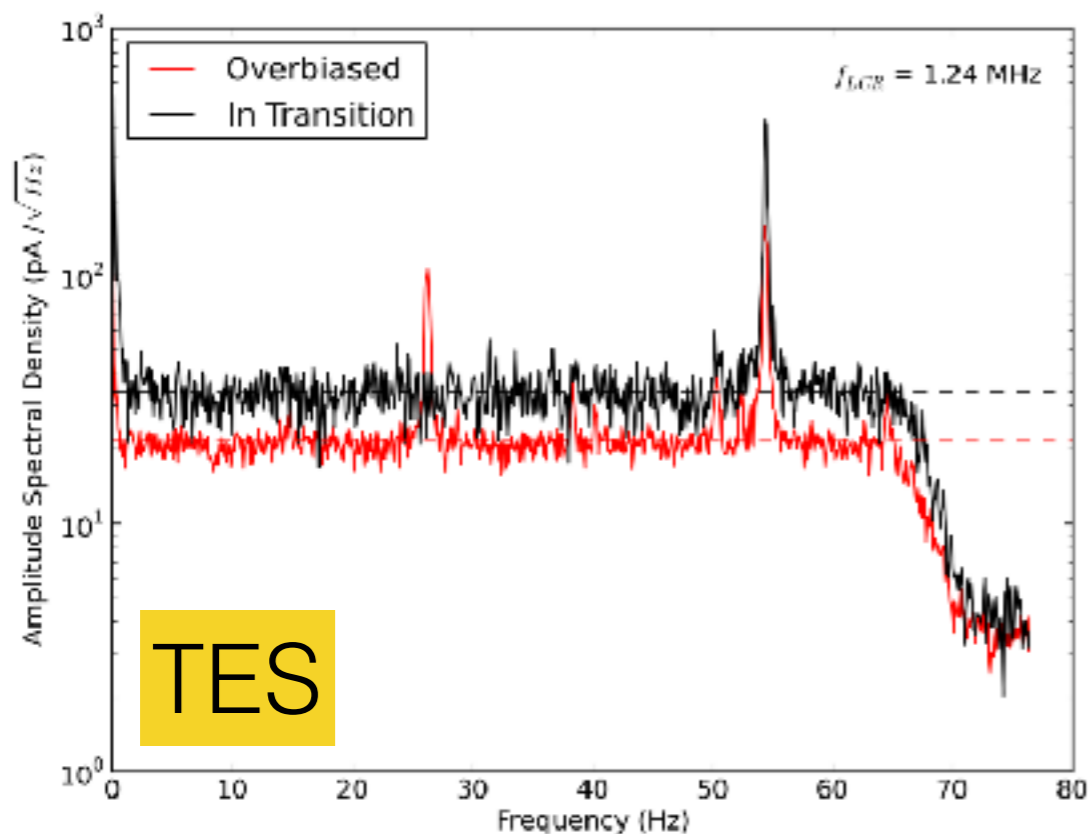
The noise RMS  $w_1$  [ $\mu$ K/ $\sqrt$ Hz], is the white noise level value, or the median value of the noise power spectral density, *referenced to a temperature scale.*



# The light: patterns



Example of noise power spectra for two different detectors, in units of  $y_{\text{det}}$  vs. frequency



The noise RMS  $w_1$  [ $\mu\text{K}/\sqrt{\text{Hz}}$ ], is the Noise Equivalent Temperature (NET)

Measurement / detector units are usually Voltage, frequency etc. as a function of time (or freq. by FFT). Let's call these units  $y_{\text{det}}$ .

1. Convert this to NEP or noise equivalent power.
2. Convert to NET by using a *Black-Body-Jacobian*

# The light: patterns

Black-Body intensity  $I$  has dimensions of W/Sr/m<sup>2</sup>/Hz

$$I(\nu, T) = \frac{2h\nu^3}{c^2} \frac{1}{e^{\frac{h\nu}{kT}} - 1},$$

1. Conversion to NEP or noise equivalent power:

Measure  $y_{\text{det}}$  for varying incident BB power and obtain the responsivity, as a function of frequency (or FFT of time streams)

$$\mathcal{R} = \frac{\delta y_{\text{det}}}{\delta P} \rightarrow P_{\text{det}}(f) = \mathcal{R}(f) y_{\text{det}}(f)$$

Spectral density of this power is the NEP

2. Conversion to NET by using a *Black-Body-Jacobian*

$$P(T) = \int d\nu dA d\Omega (W(\nu) \cdot I(\nu, T))$$

$$\rightarrow J = \frac{\partial P}{\partial T}$$

$$w_1 = \text{NET} = \text{NEP} J^{-1}$$

# The light: patterns

NEP and photon noise limit

**Noise Equivalent Power:** input signal power that produces  $\text{SNR} = 1$  at the output of a detector, given data-signaling rate / modulation frequency, and effective noise bandwidth.

Thus it is the minimum detectable power per  $\sqrt{\text{bandwidth}}$ .

The response of a detector can vary with frequency:  $\text{NEP}(f)$



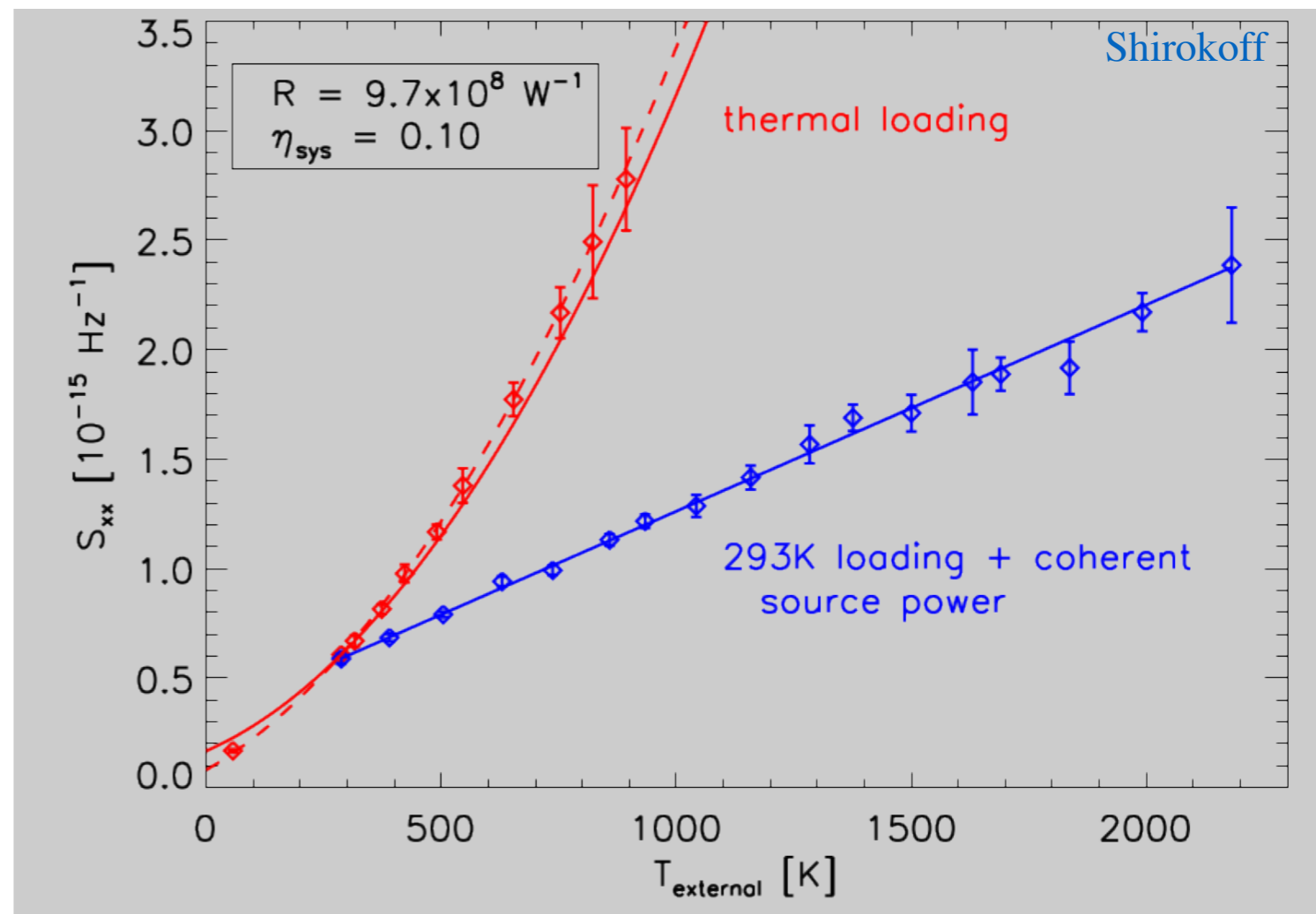
# The light: patterns

## NEP and photon noise limit

For a detector with negligible intrinsic (thermal) and readout (laboratory) noise, photon counting determines the measurement limit

$$\langle n_{\text{rms}} \rangle = \sqrt{n + n^2}$$

$h\nu \ll k_B T$        $h\nu \gg k_B T$   
Bose bunching      Poisson

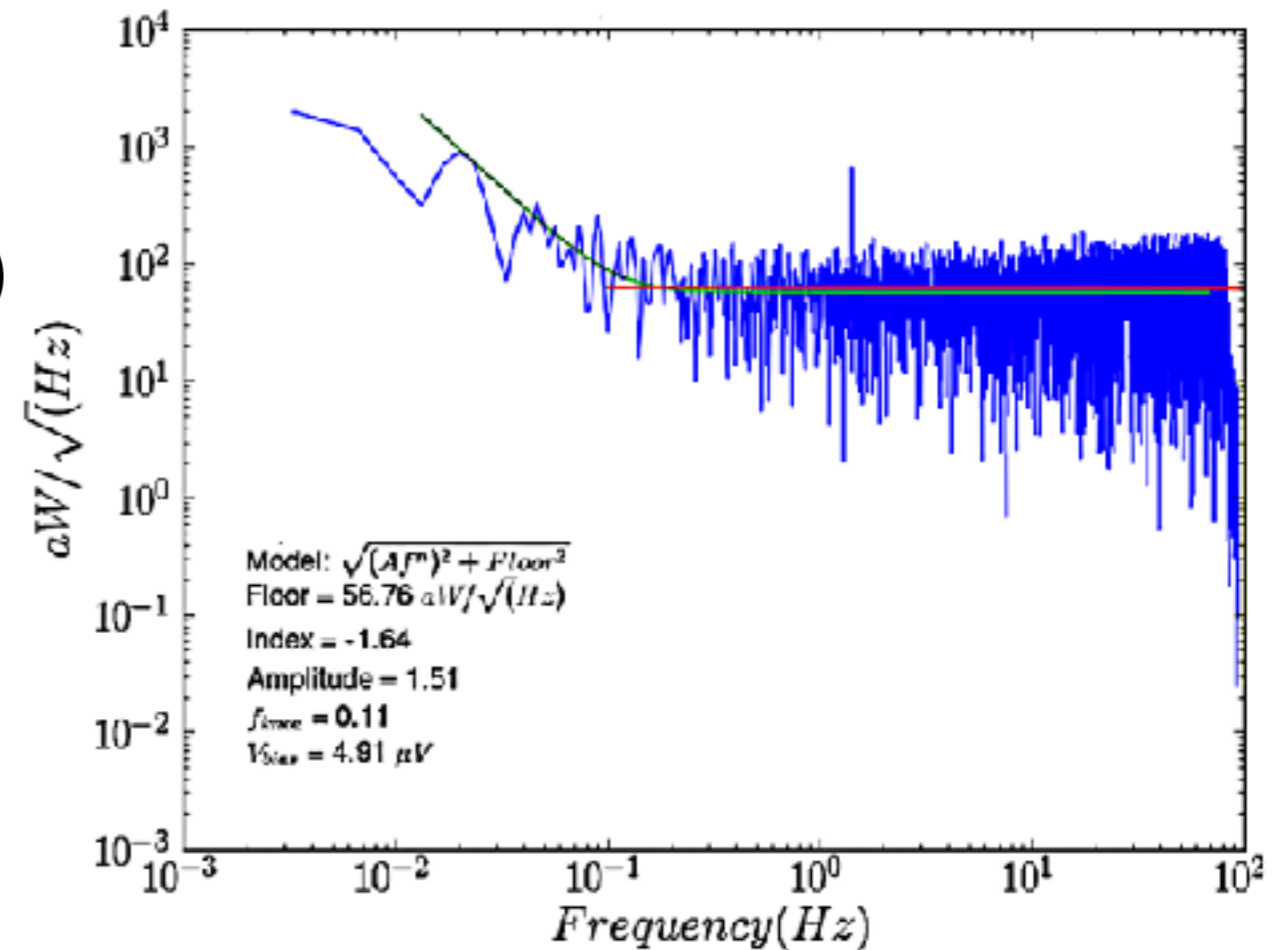


# The light: patterns

## NEP and photon noise limit

For ground based experiments  
NEP (CMB + hot optics + hot Sky)  
limit is  $\sim 50 \text{ aW}^*/\sqrt{\text{Hz}}$

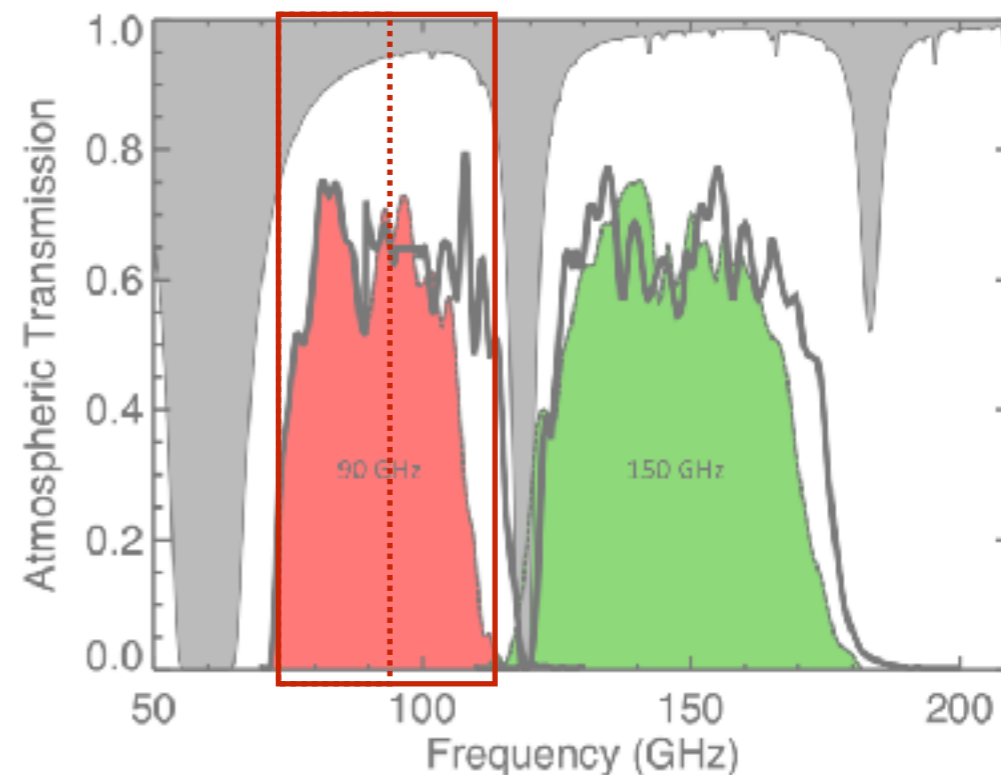
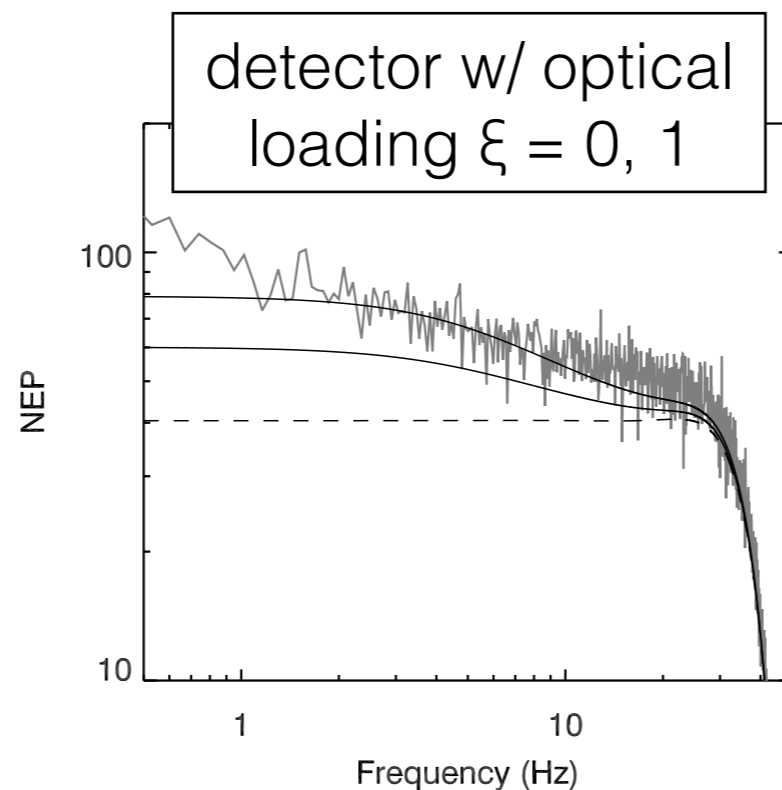
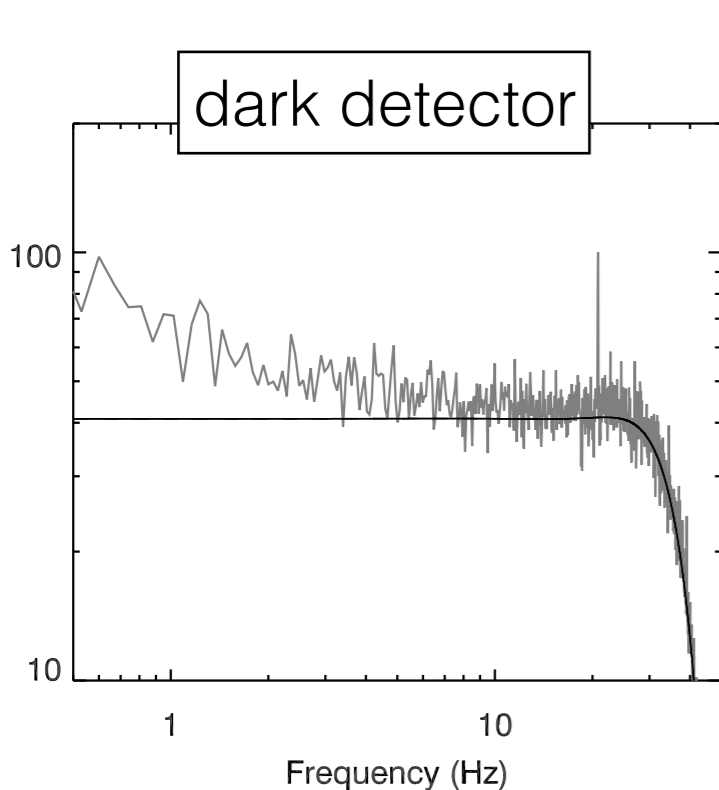
SPTpol, single detector  $< 10^2 \text{ aW}$ ,  
therefore with  $> 10^3$  detectors  
CMB can be measured.



\*1 aW =  $10^{-18}$  W

# The light: patterns

NEP and photon noise limit



$$\langle n_{\text{rms}} \rangle = \sqrt{n + n^2}$$

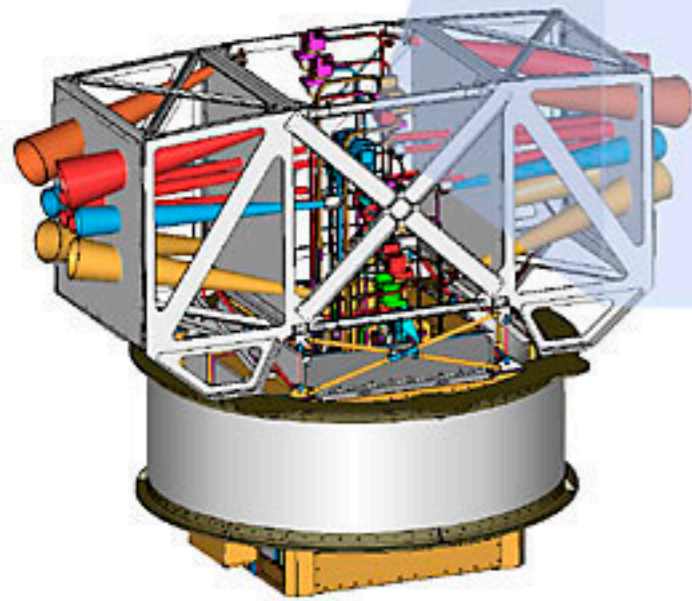
$$\text{NEP}^2 = \left( \underbrace{2h\nu_0 P_{\text{opt}}}_{\text{band-center}} + \underbrace{\frac{\xi P_{\text{opt}}^2}{\Delta\nu}}_{\text{band-width}} \right)$$

Bose-bunching- mode collection efficiency



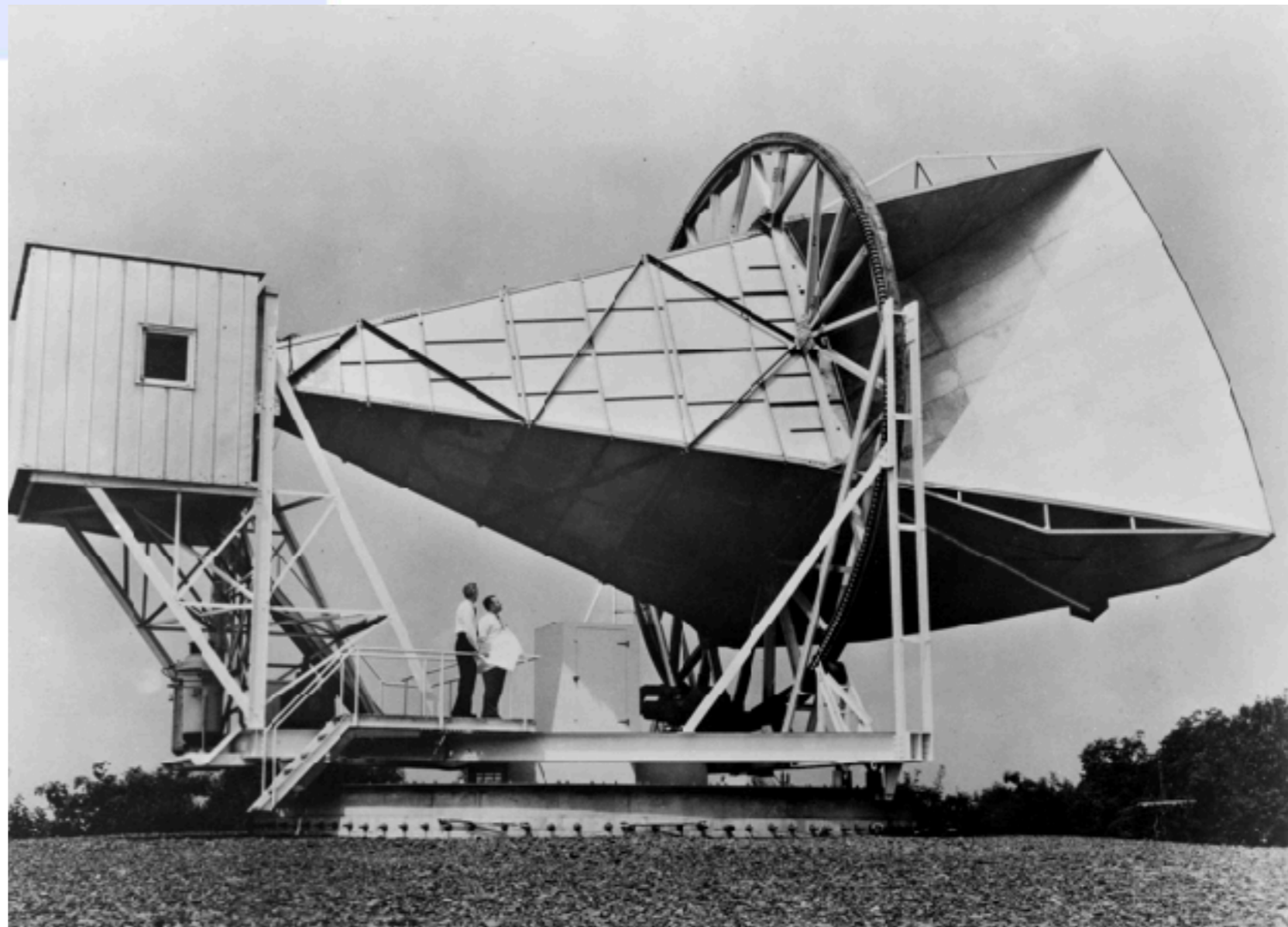
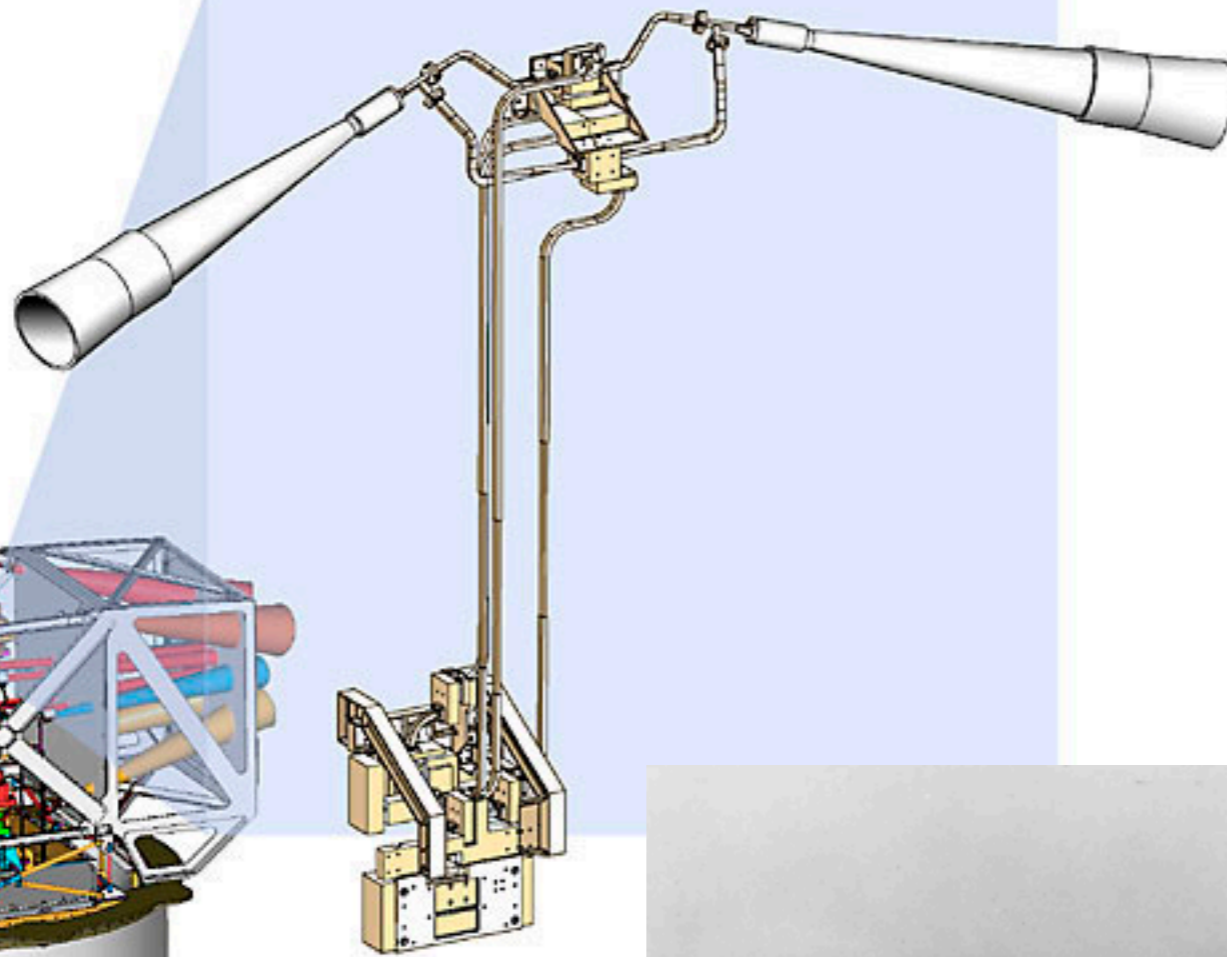
Let the light in

Let the light in



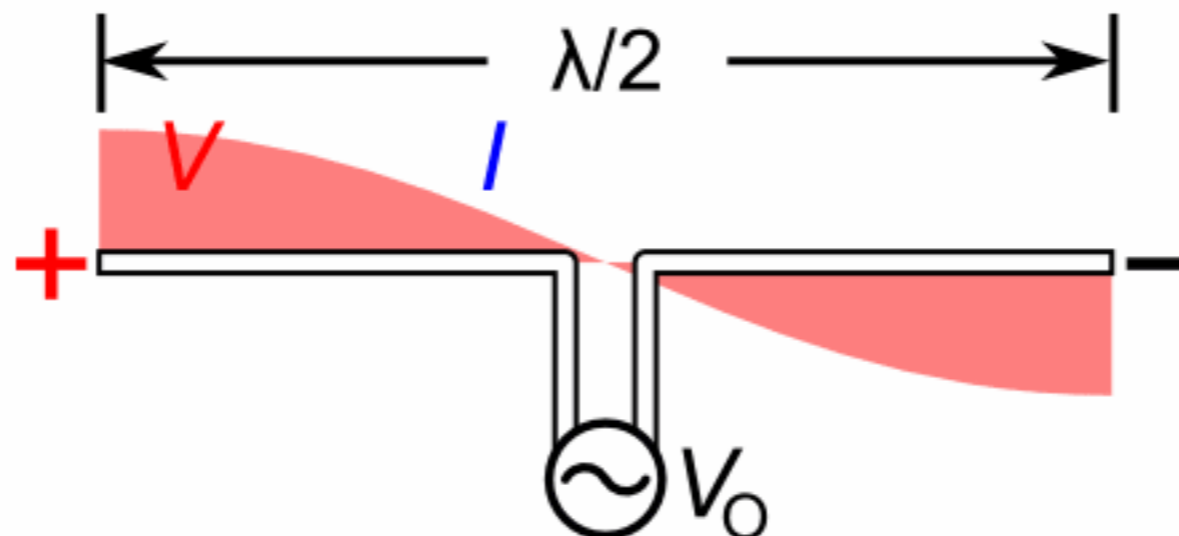
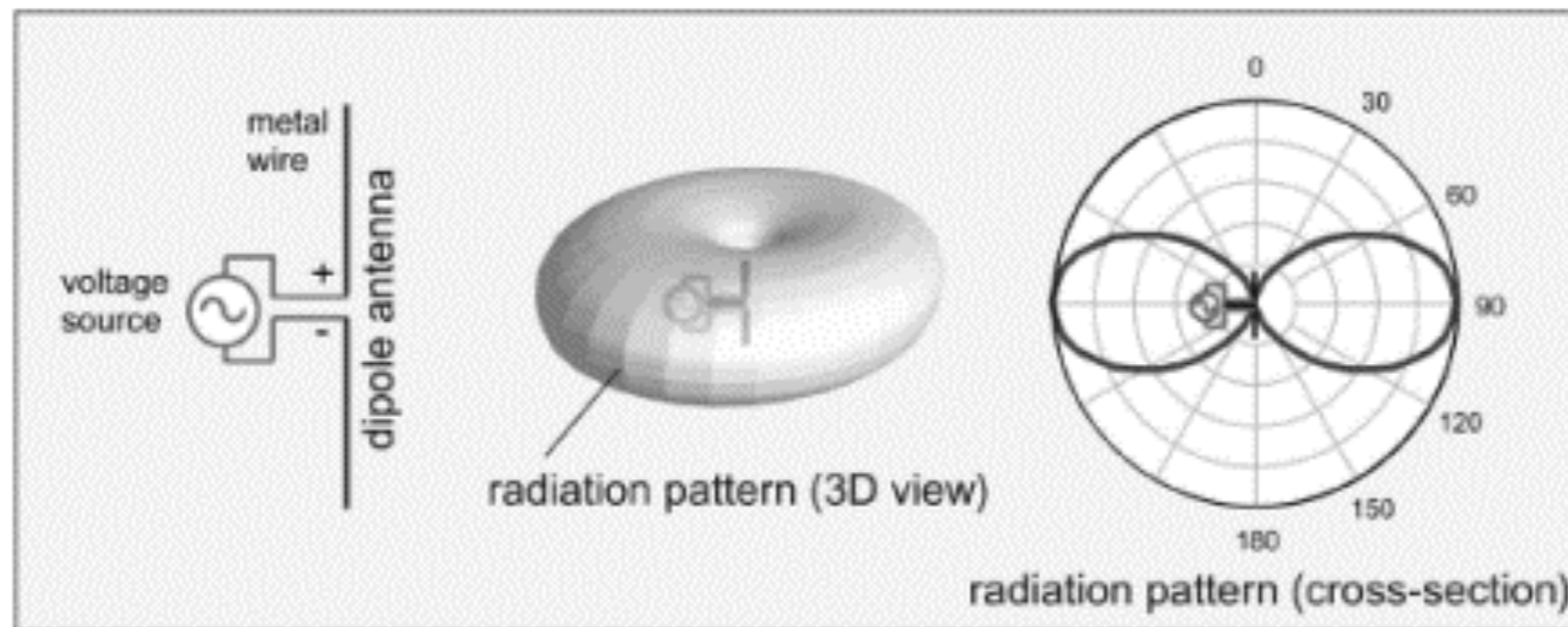
WMAP

Penzias  
& Wilson

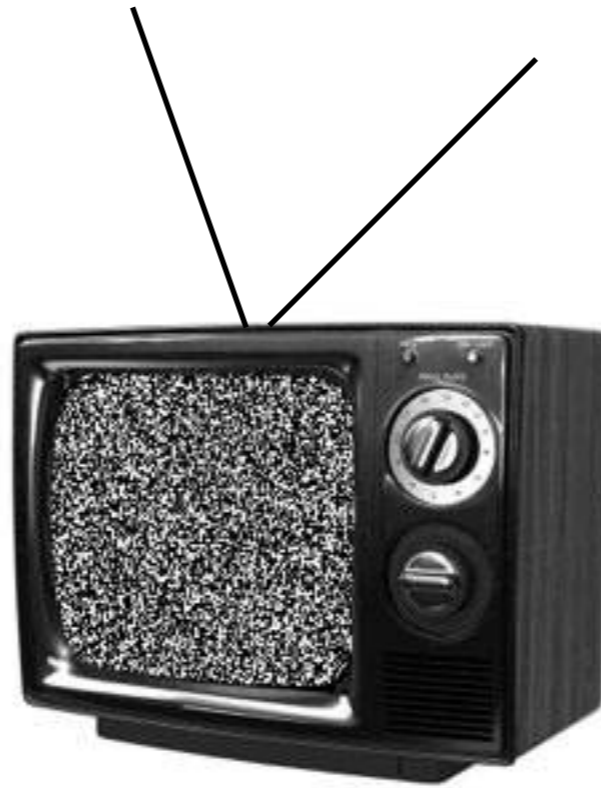


## Reciprocity theorem:

*Receive and transmit* properties of an antenna are identical.  
Radiation pattern in transmit mode = pattern in the receive mode.



Let the light in

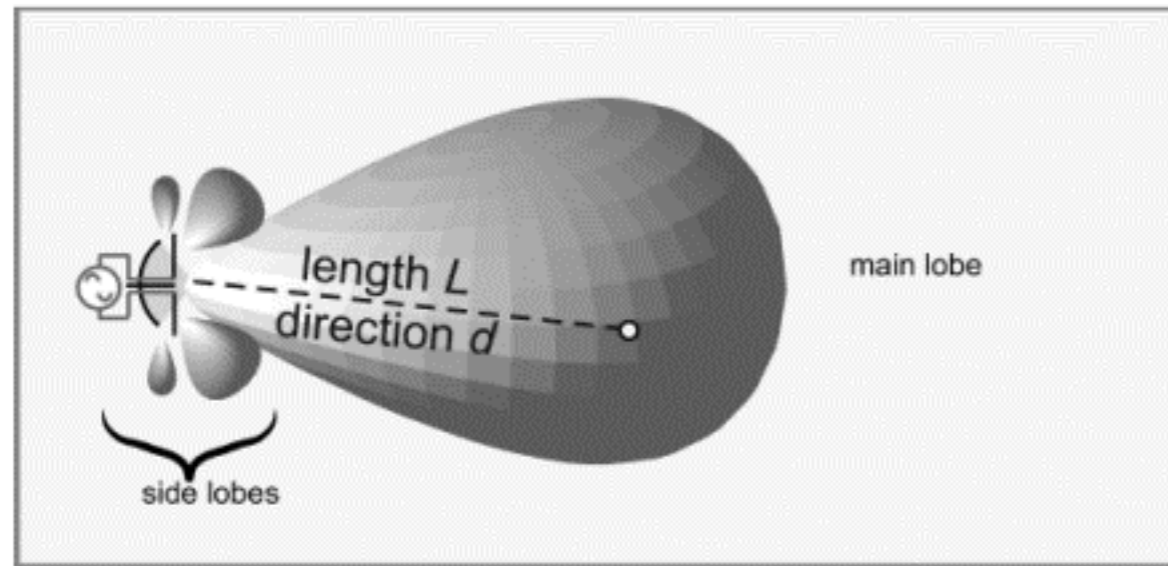


~1% of TV noise is the CMB

But we can do better using horns ...



Let the light in

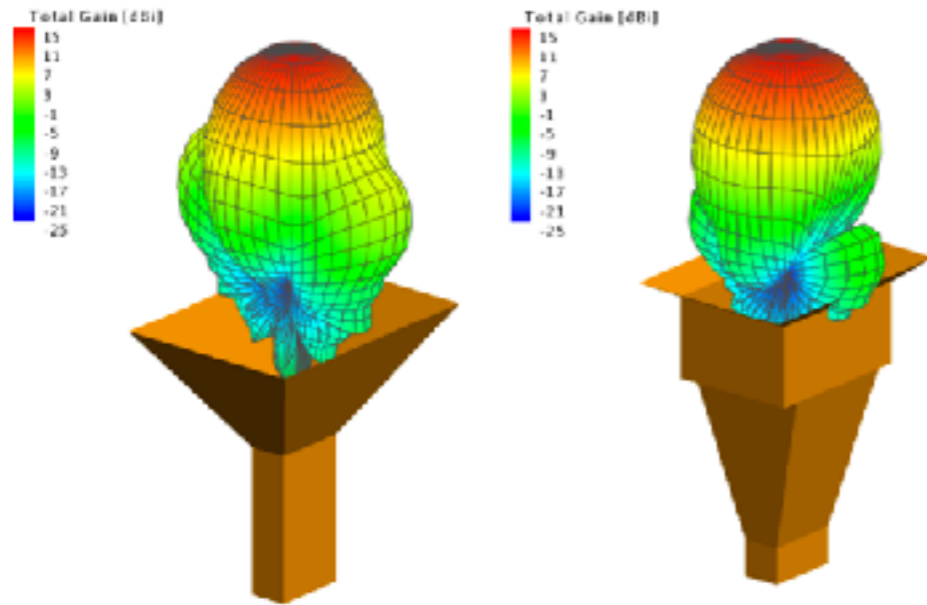


Shape the lobes, so as to focus the CMB from the sky and not pick up terrestrial junk

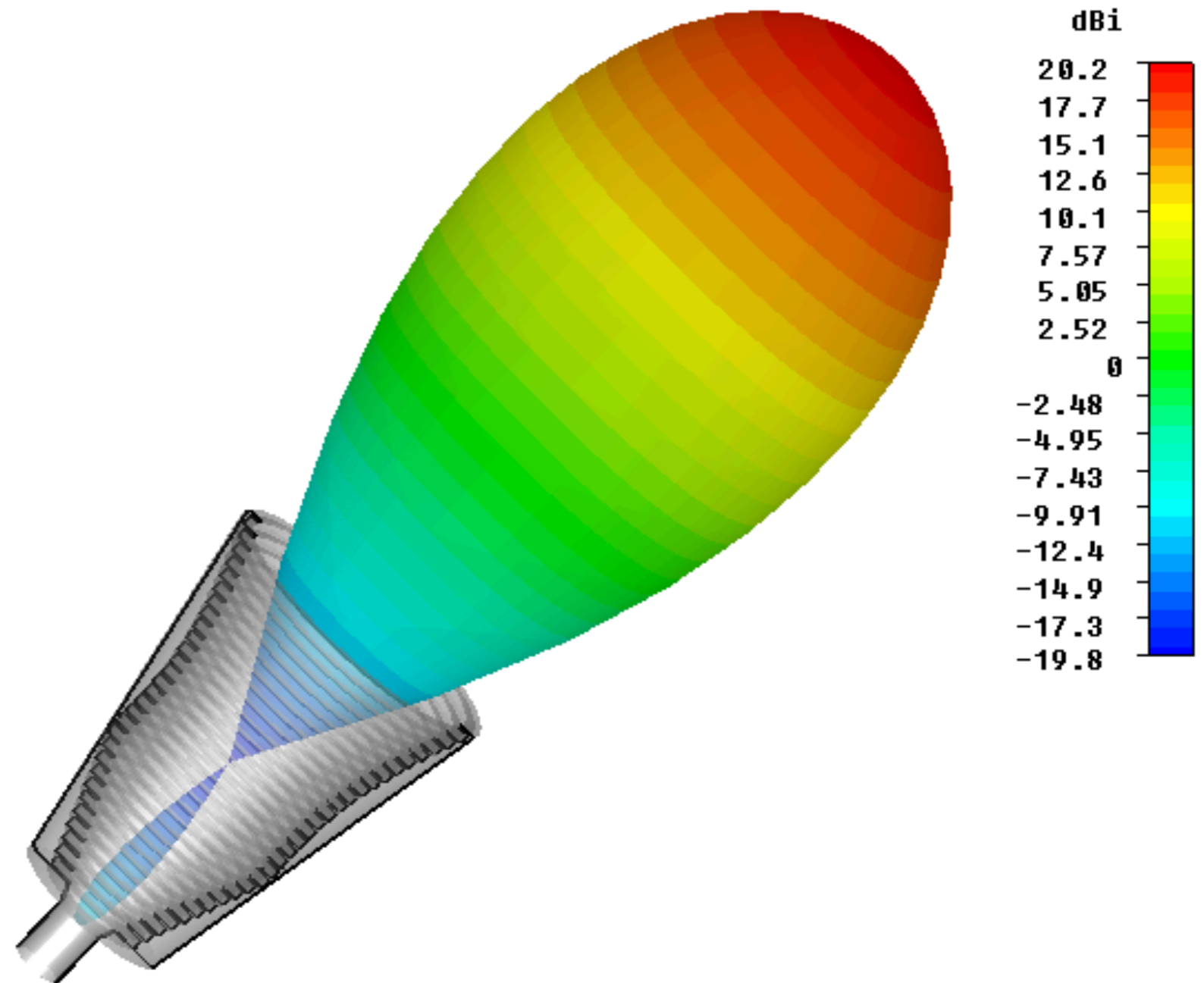
Horsing around, but more seriously.....



Let the light in

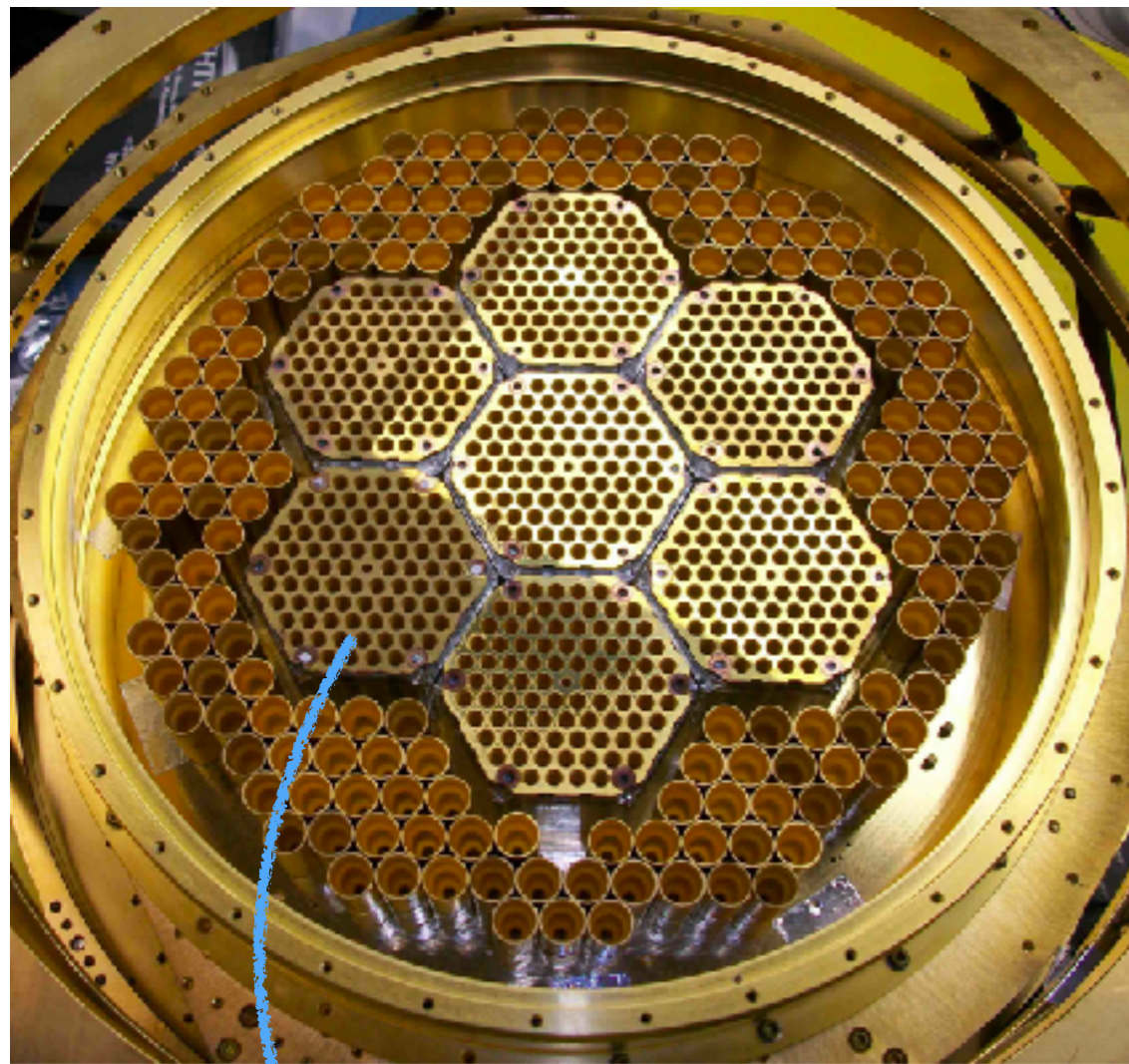


A well defined corrugated horn antennae can couple the entire power in the main lobe

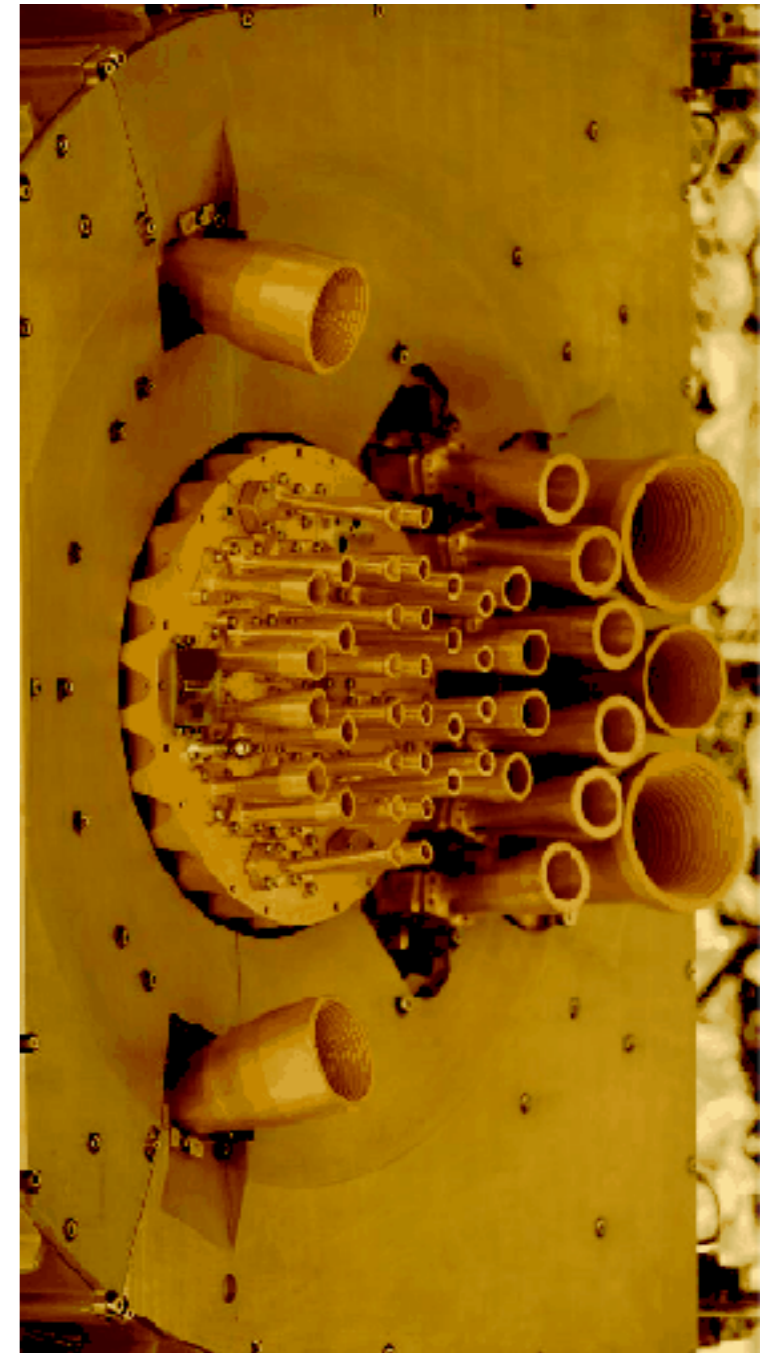




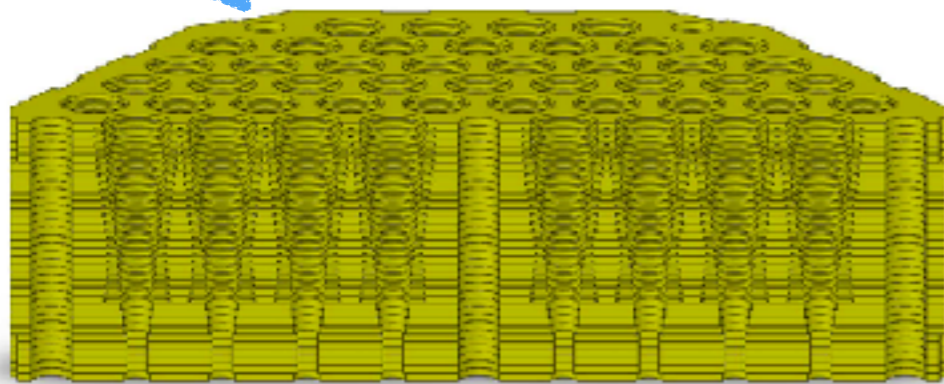
Let the light in



Planck →



← SPTPoI

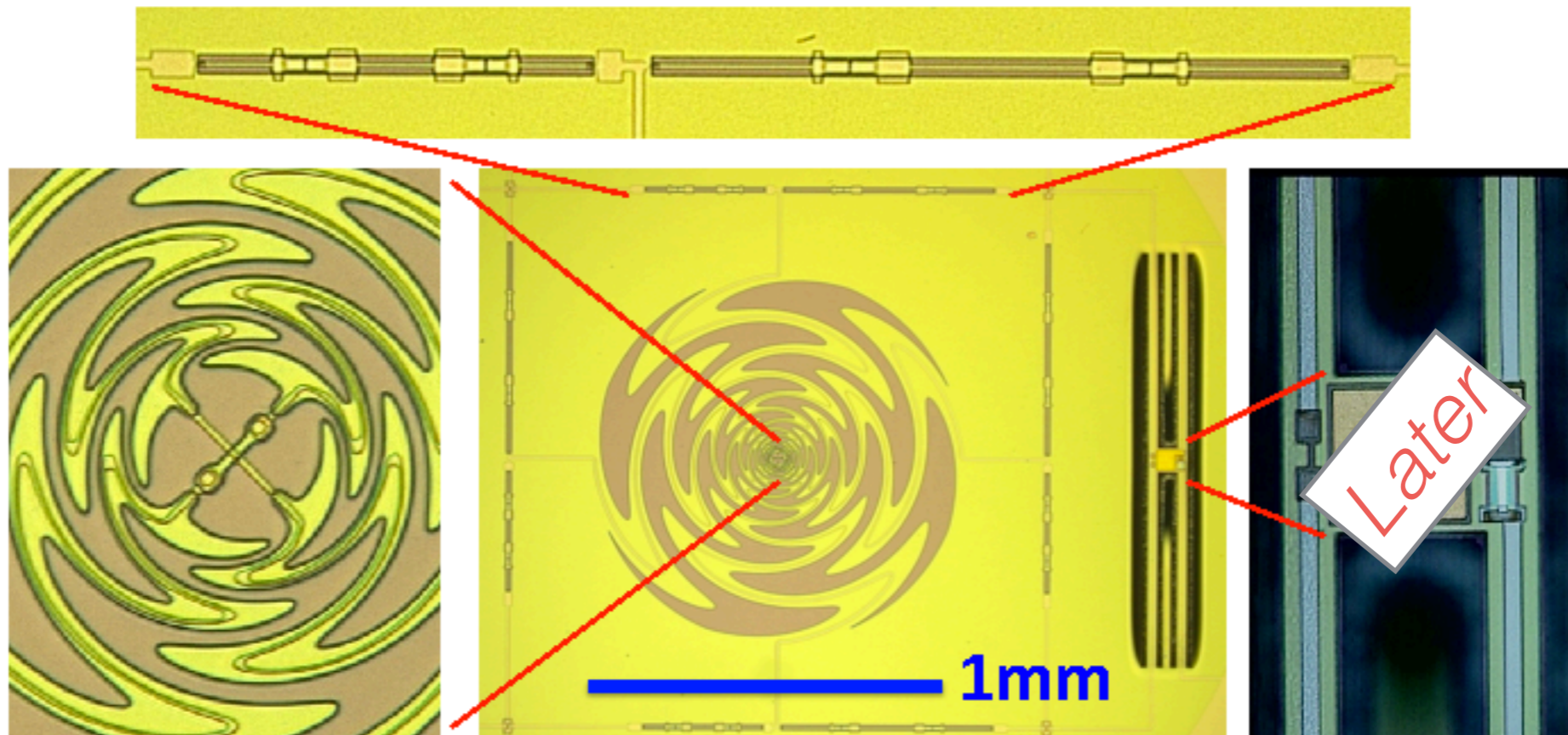




# Let the light in

A horn has one well defined length-scale ( $L_H$ ), and thus the response or gain will drop for all  $\nu < c/L_H$

But ideally we want to pick up “equally” at large bandwidths and then chop it into bands that we care for

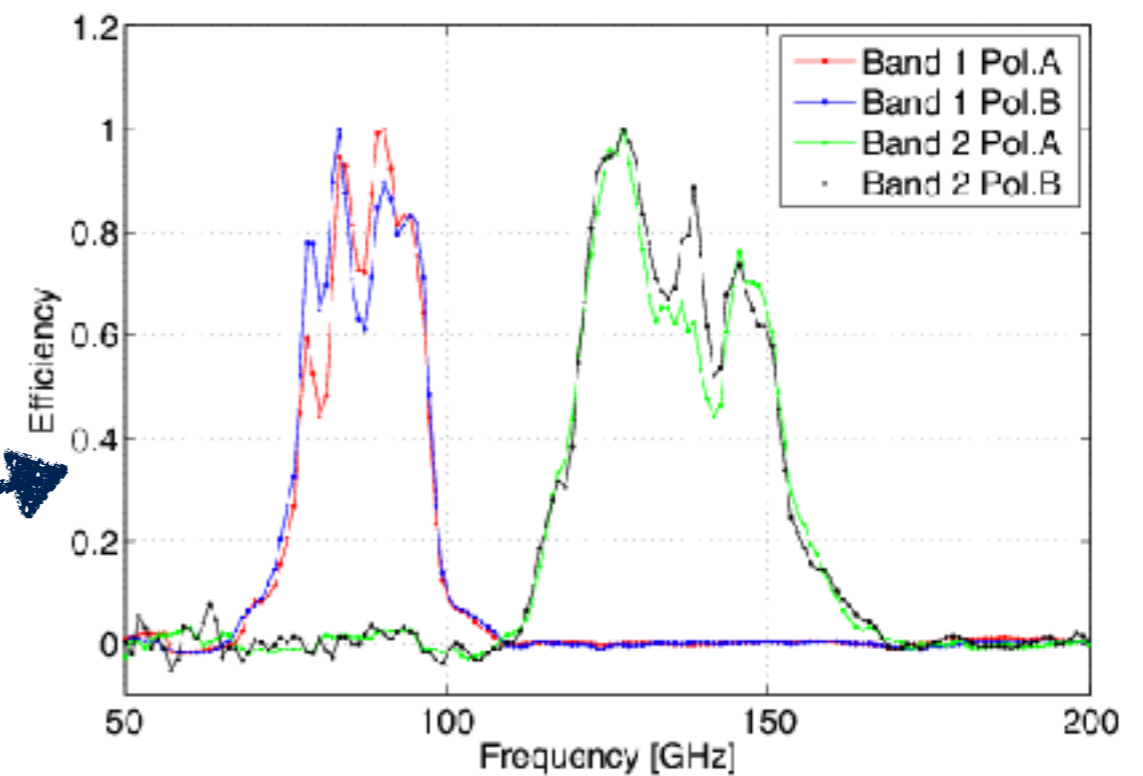
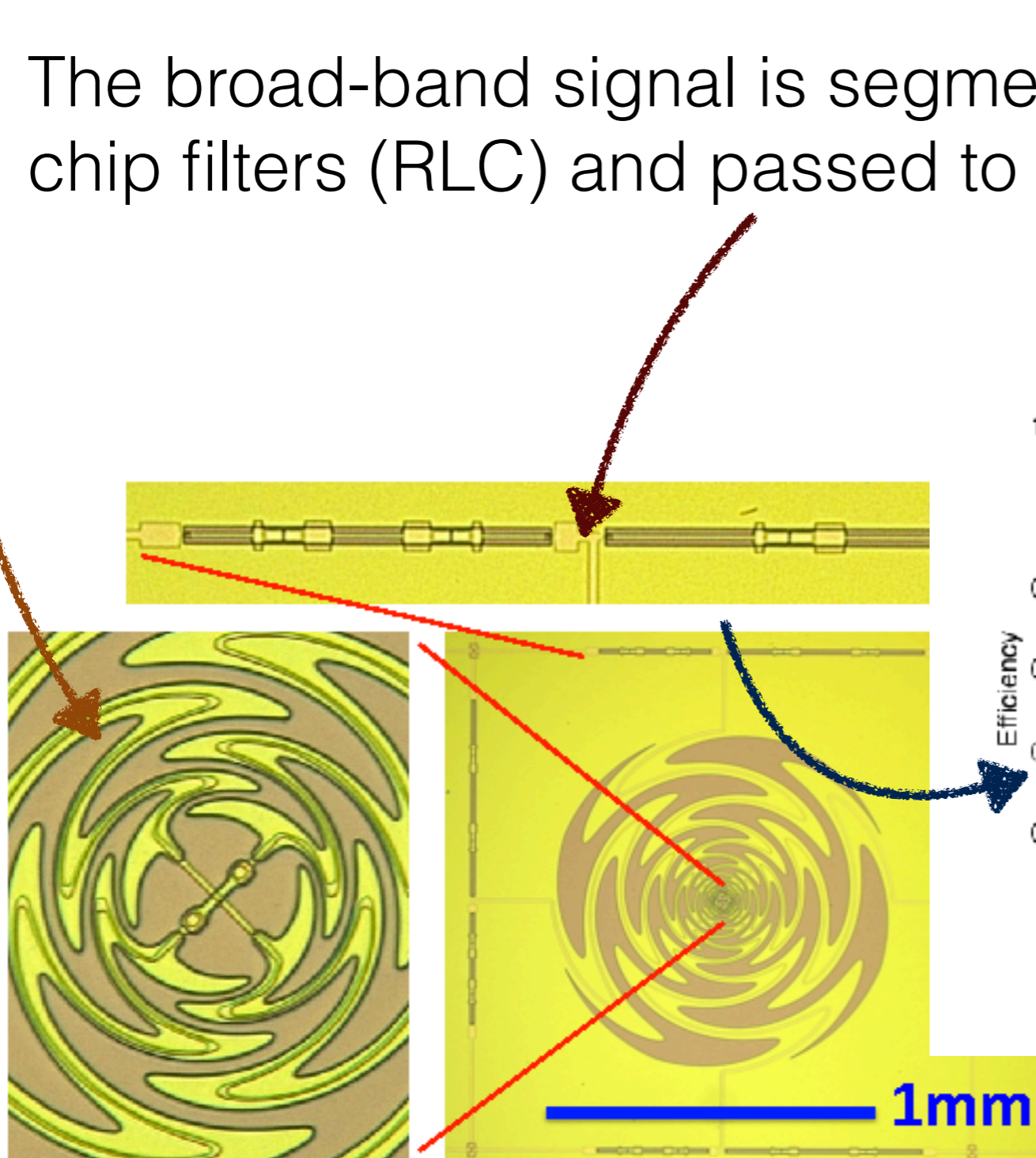




# Let the light in

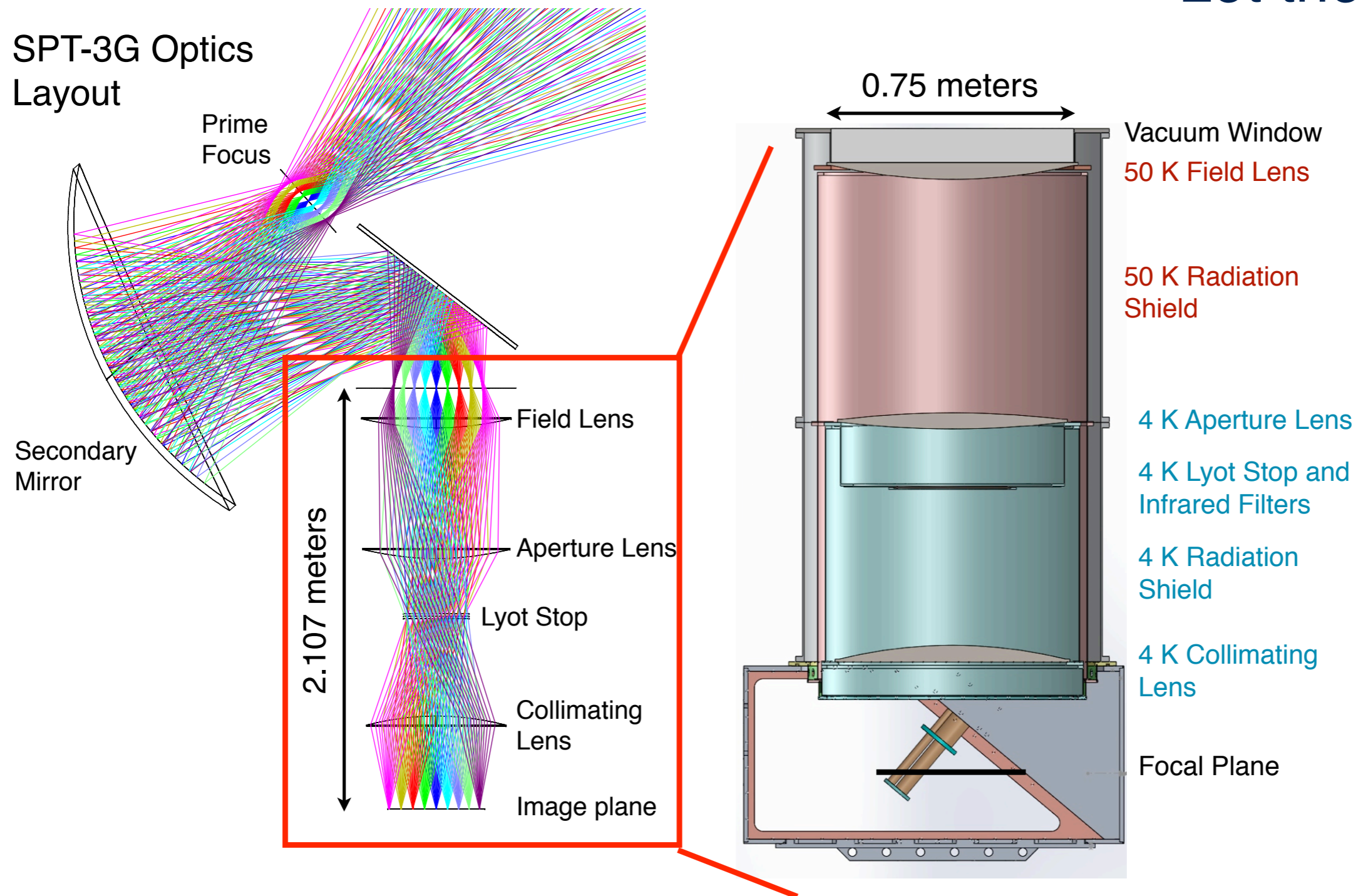
Log periodic / fractal shape is “coherent” across decades

The broad-band signal is segmented by lumped-element on-chip filters (RLC) and passed to detectors for measurement



SPT3G, PB2 etc.

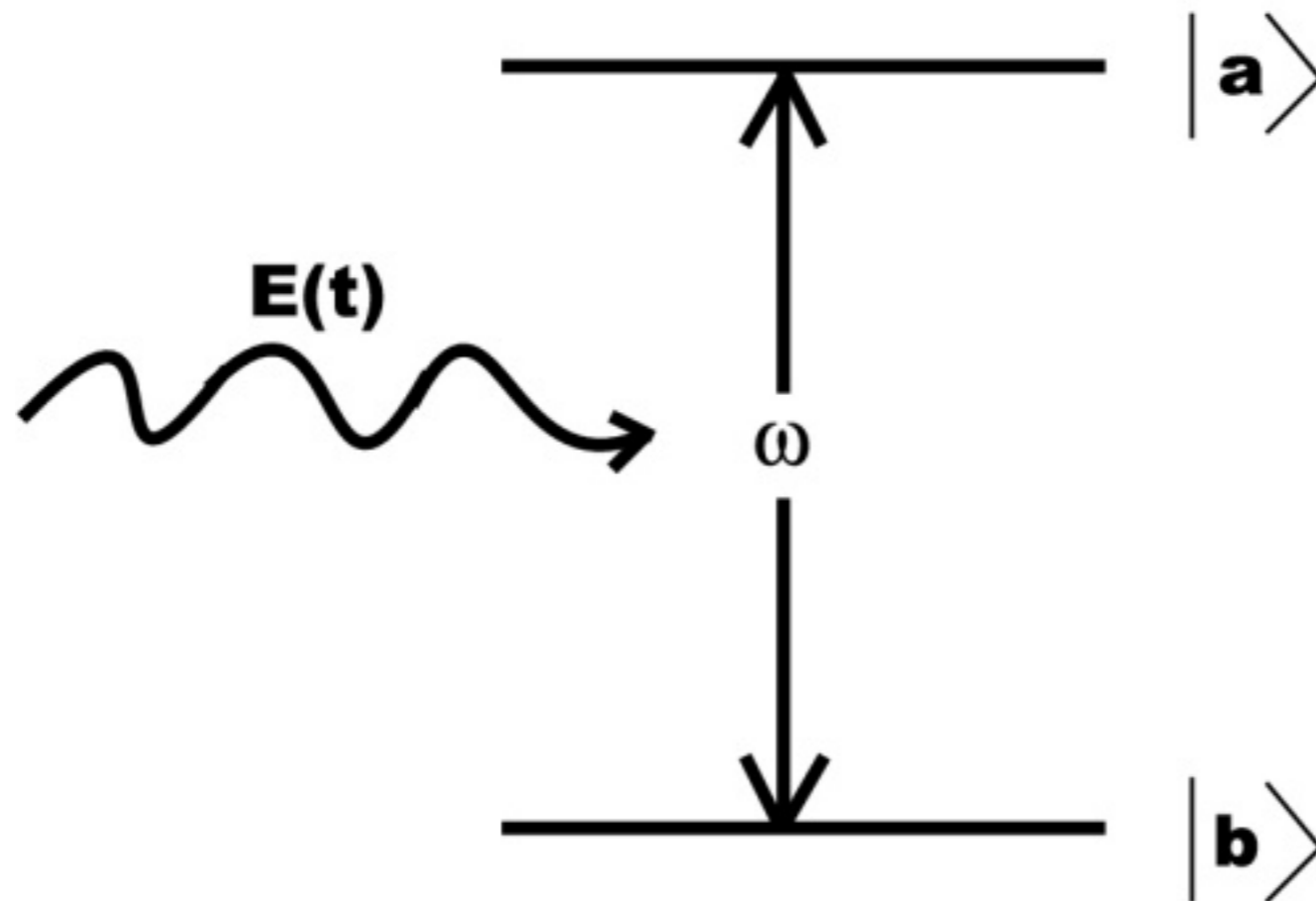
# Let the light in



Various optical elements, lenses, collimators, filters etc. are non-trivial at mm waves, especially for big (3G-like) instruments. R&D in microwave material science is happening, and more effort is required

Seeing the light

# Seeing the light



At the heart of all CMB experiments, is quantum excitation

We will discuss:

WMAP receivers, SPT detectors and one CMBS4 technology



# Seeing the light: WMAP

WMAP looked at the difference in signals from two horns

The “radio” signals are amplified by HEMT amplifiers

The amplified power is measured with diodes

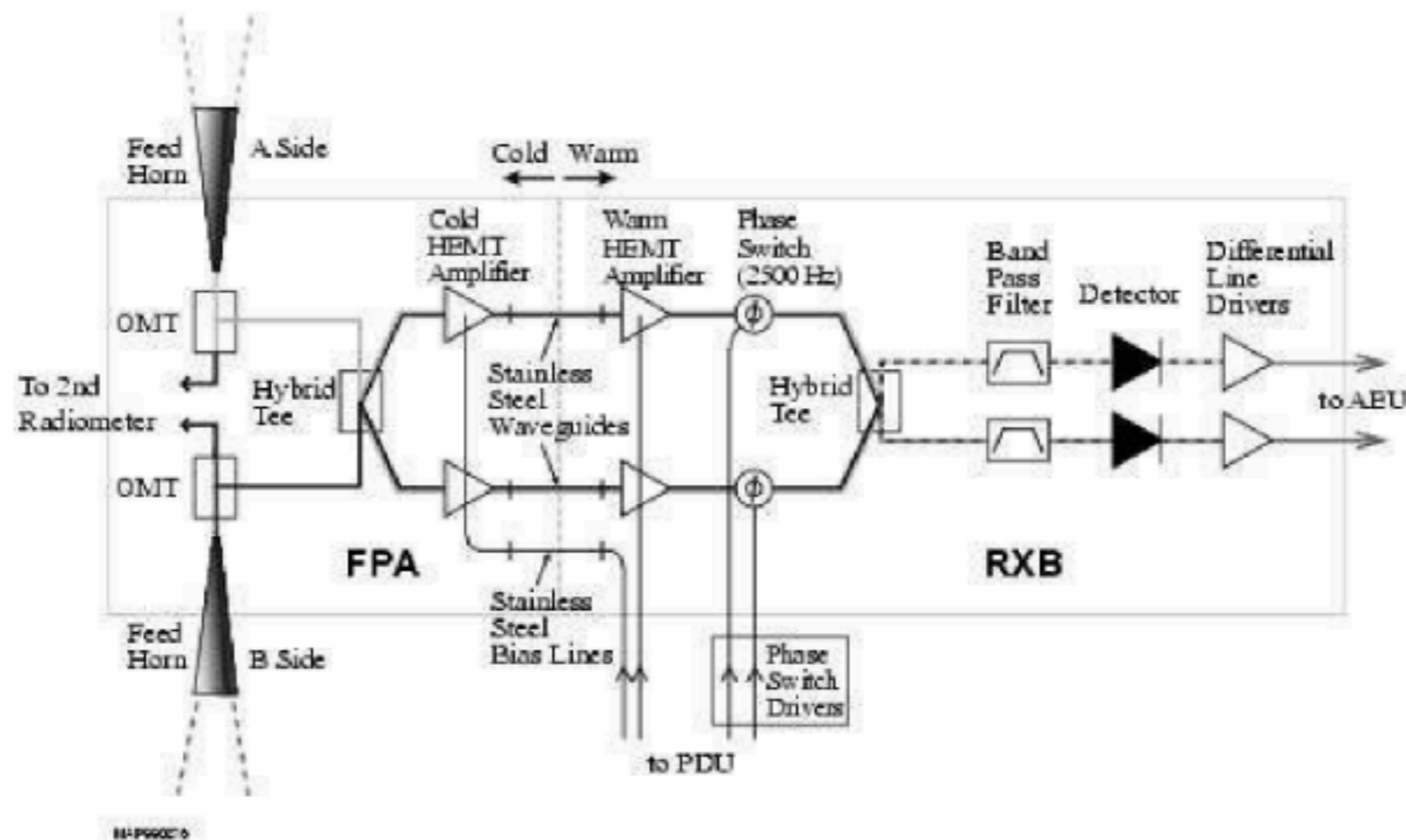
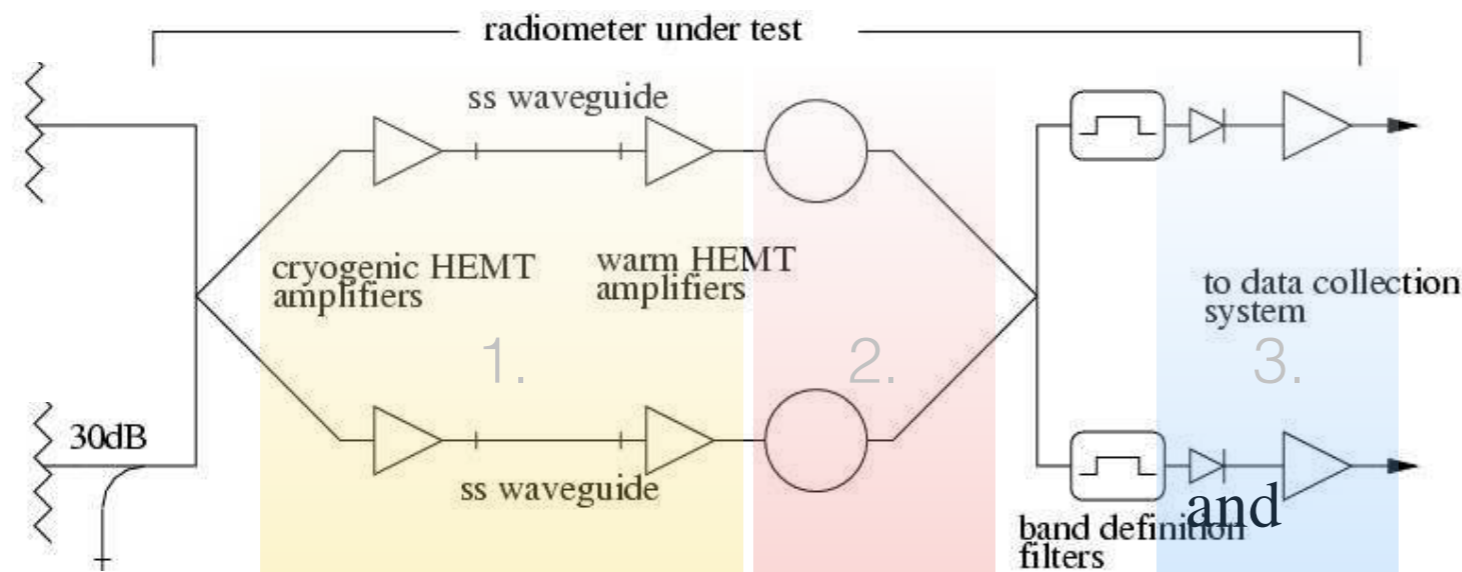


FIG. 1.— Layout of an individual MAP radiometer. Components on the cold (left) side of the stainless steel waveguides are located in the FPA, and are passively cooled to 90 K in flight.

# Seeing the light: WMAP

$$\sigma_A^2 \propto k_b T_A \Delta\nu$$

The outputs of the two detectors then become



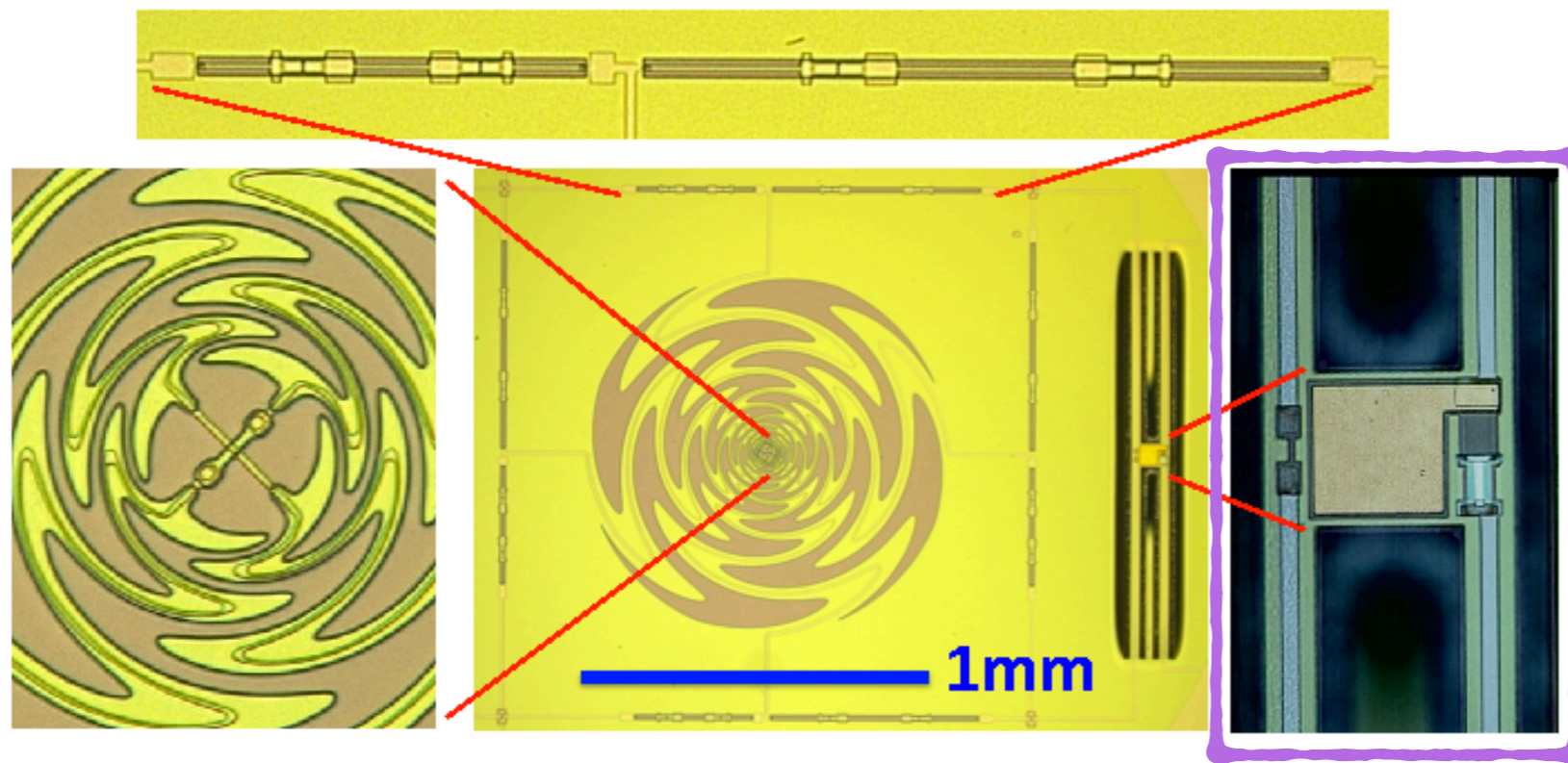
$$V_l = \frac{s}{2} \left\{ \left( \frac{A^2 + B^2}{2} + n_1^2 \right) g_1^2(t) + \left( \frac{A^2 + B^2}{2} + n_2^2 \right) g_2^2(t) \mp (A^2 - B^2) g_1(t) g_2(t) \right\}$$

$$\sigma_B^2 \propto k_b T_B \Delta\nu$$

$$V_r = \frac{s}{2} \left\{ \left( \frac{A^2 + B^2}{2} + n_1^2 \right) g_1^2(t) + \left( \frac{A^2 + B^2}{2} + n_2^2 \right) g_2^2(t) \pm (A^2 - B^2) g_1(t) g_2(t) \right\}$$

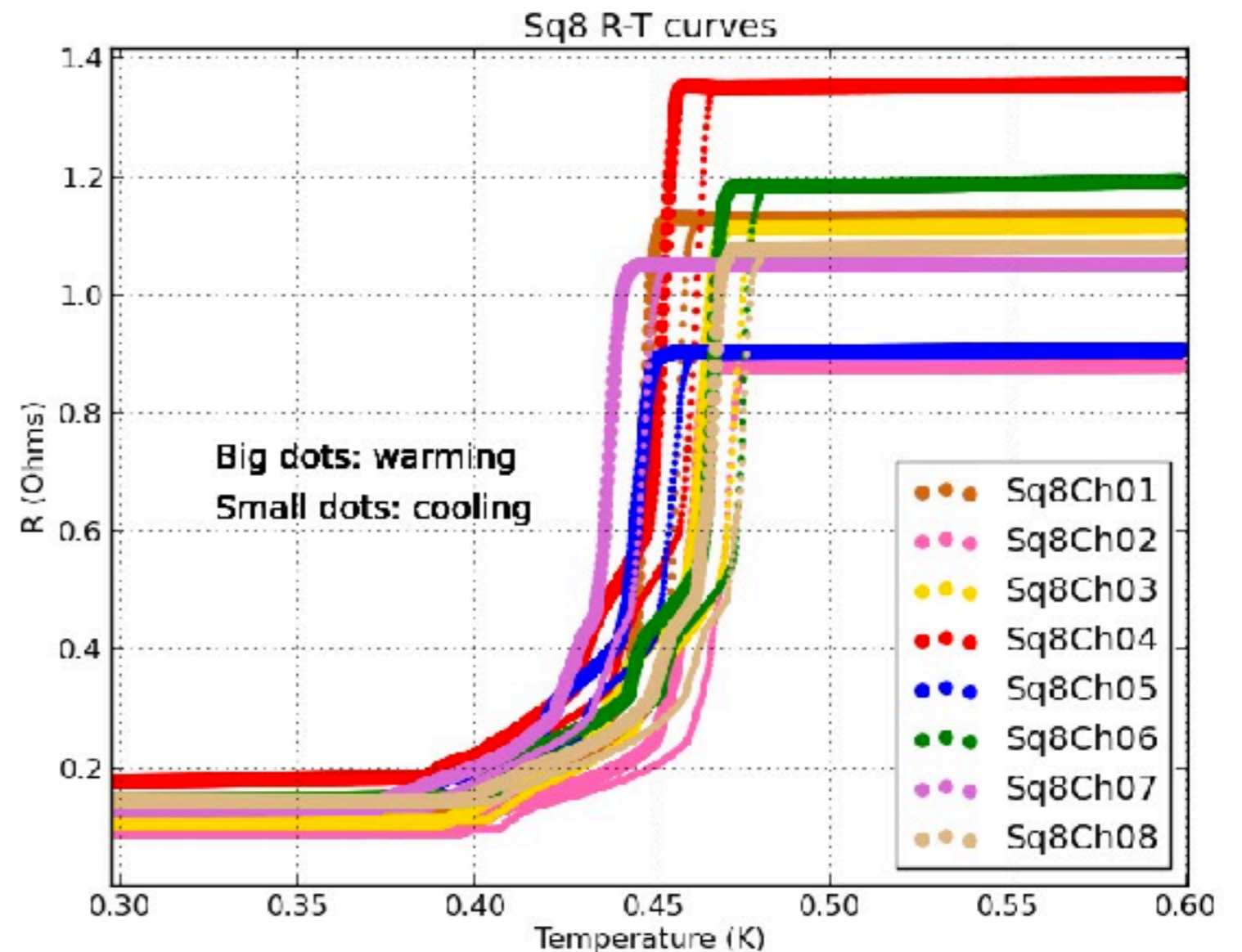
1. Amplifiers (HEMT) with gain  $g_{1/2}$  and noise  $n_{1/2}$
2. Phase offsets
3. Diodes responds linearly to input power
4. Answer  $\Rightarrow V_r - V_l$

# Seeing the light: SPT

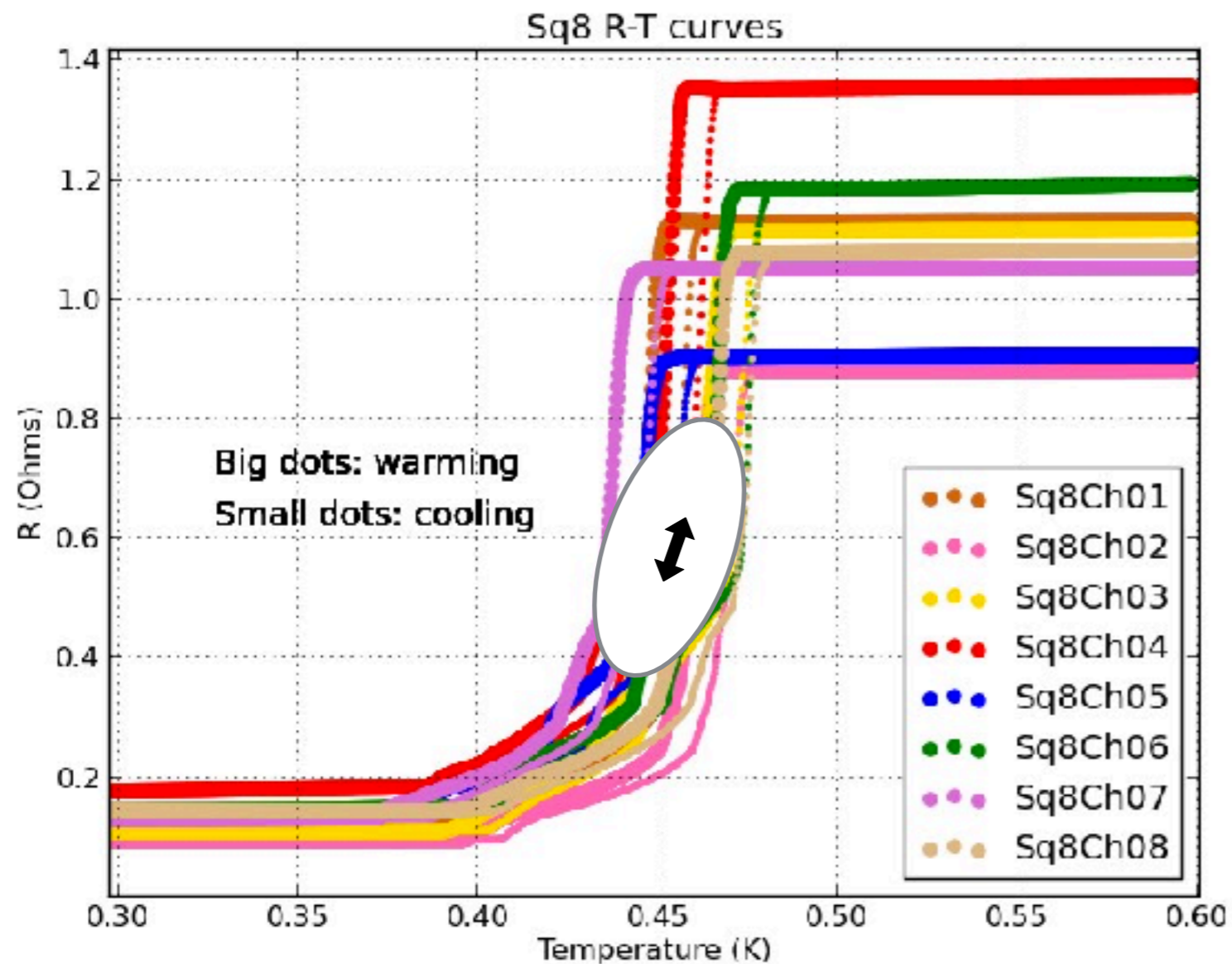


Transition edge sensors (TESs)

Resistance change of a superconductor as you heat it up a bit.



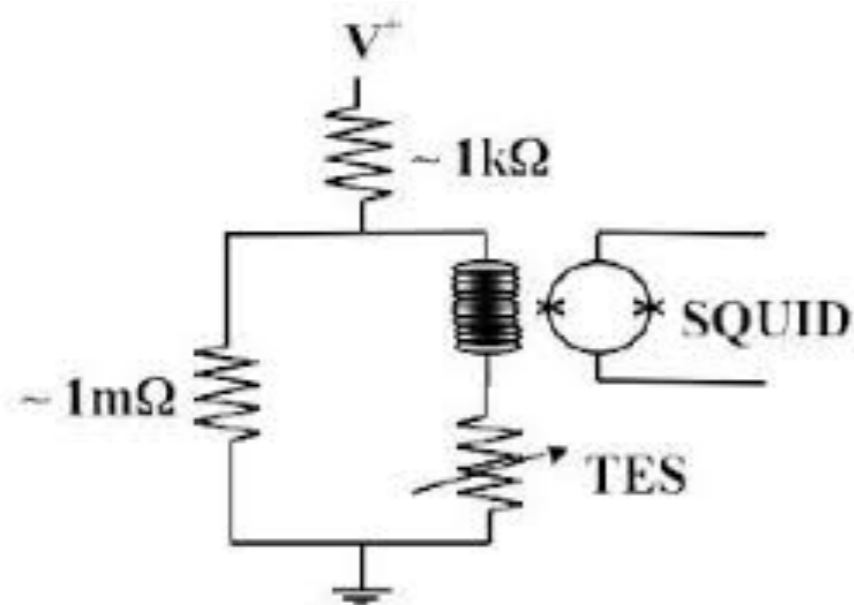




## Seeing the light: SPT

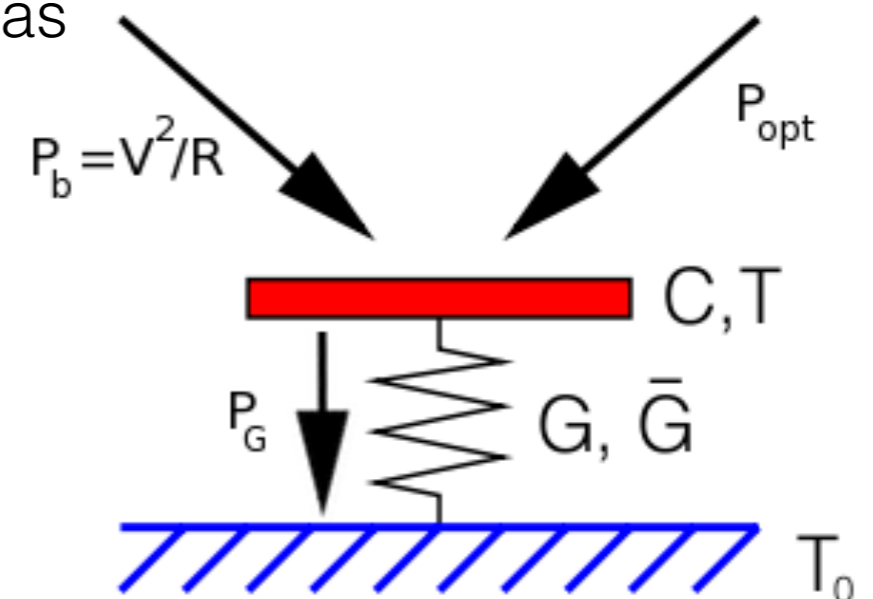
Voltage-bias the detector to sit on transition:

Small  $dT \rightarrow$  large  $dR$   
large  $dR \rightarrow$  large  $dI$



Joule-heating  
from Voltage-bias

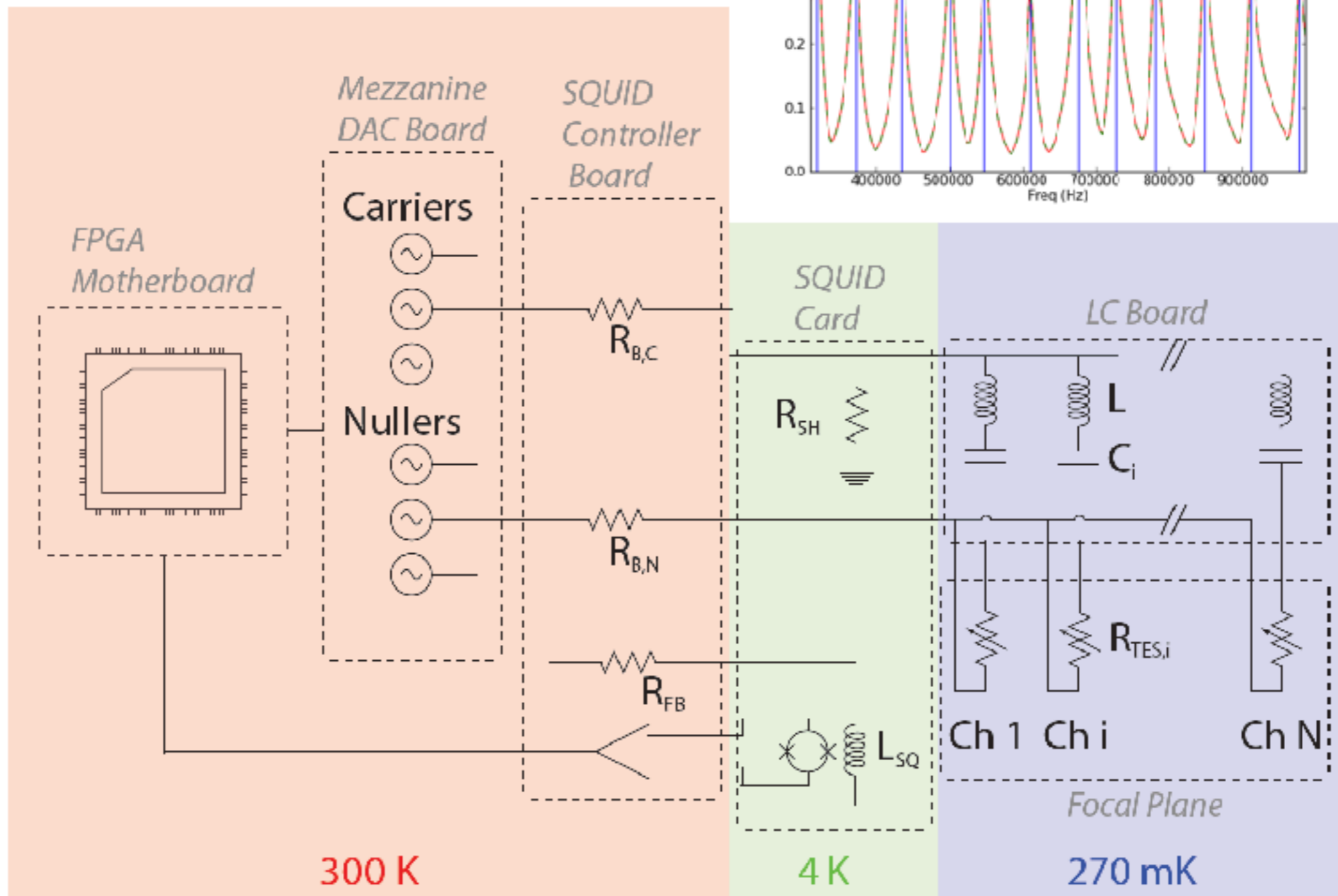
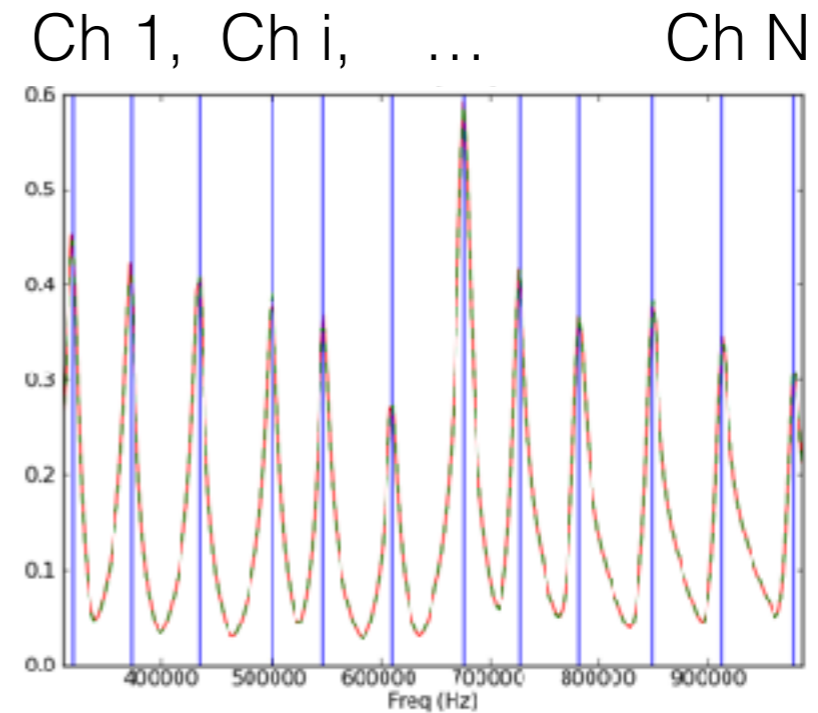
Optical-heating  
from CMB, Sky etc.



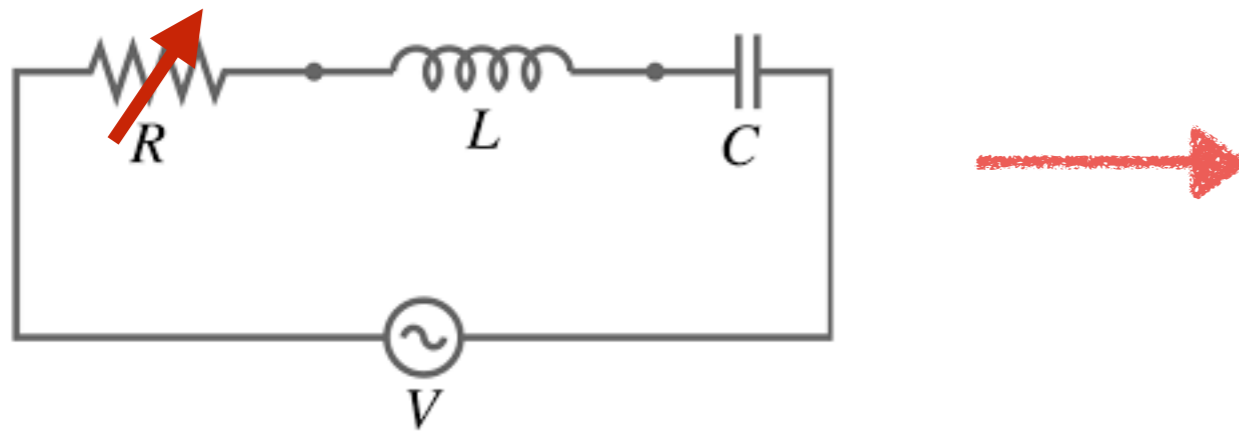


# Seeing the light: SPT

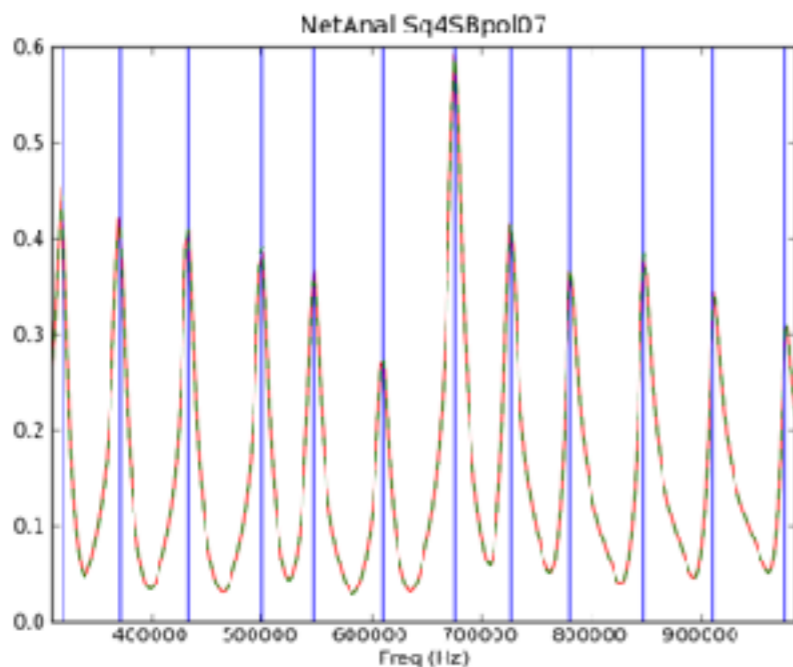
We have to read thousands of TESs!  
Clever but complicated



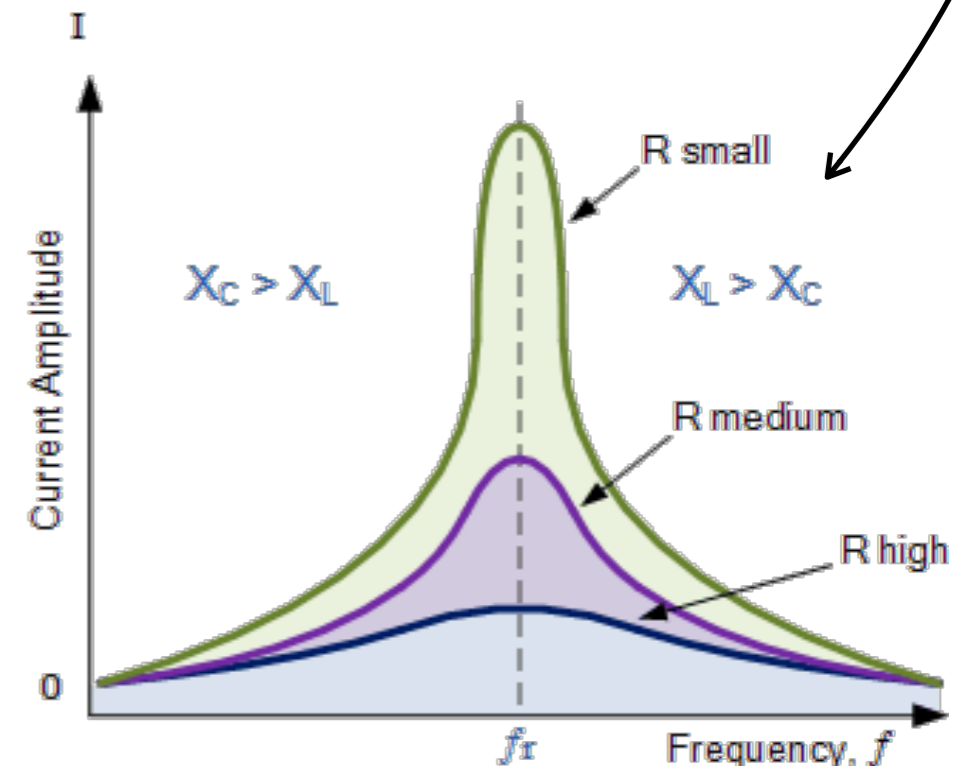
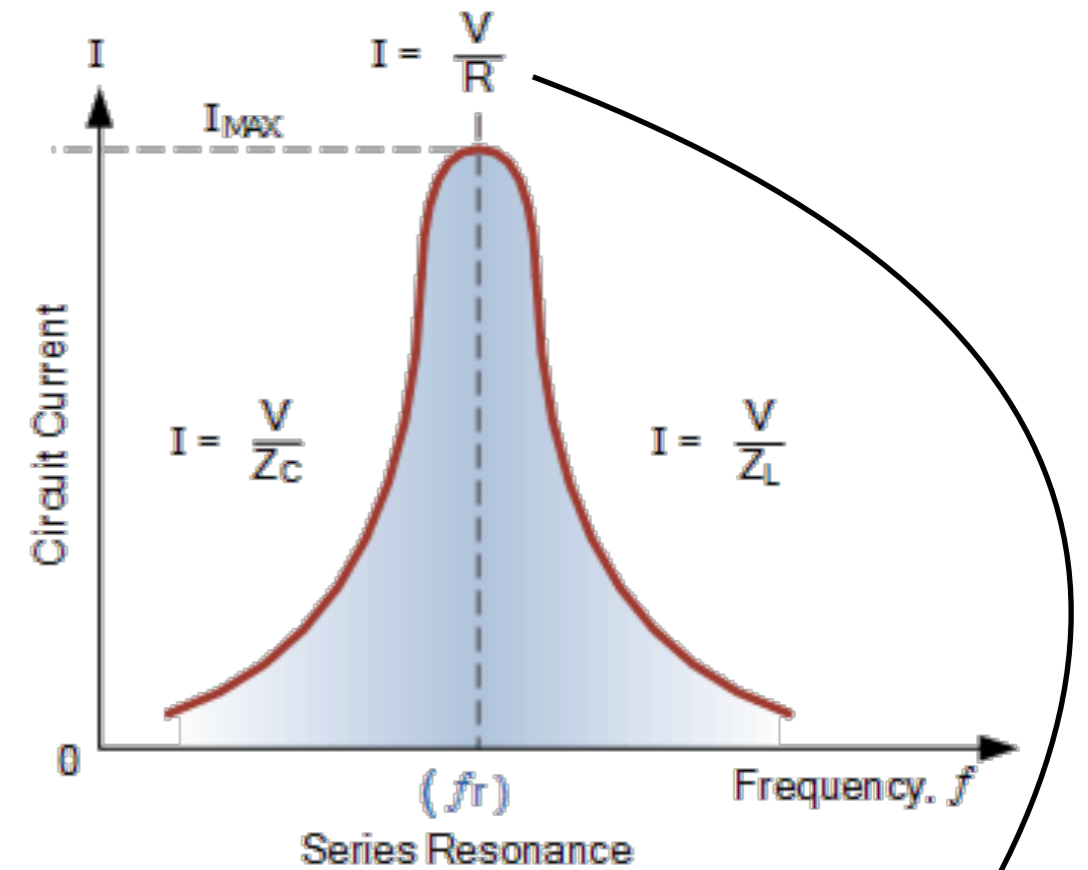
We have to read thousands of TESs! Clever but complicated



Couple a TES to a LC system.  
Amplitude is modulated by  $R(T)$ ,  
and each TES gets a resonant  $f_r$ ....  
therefore we multiplex

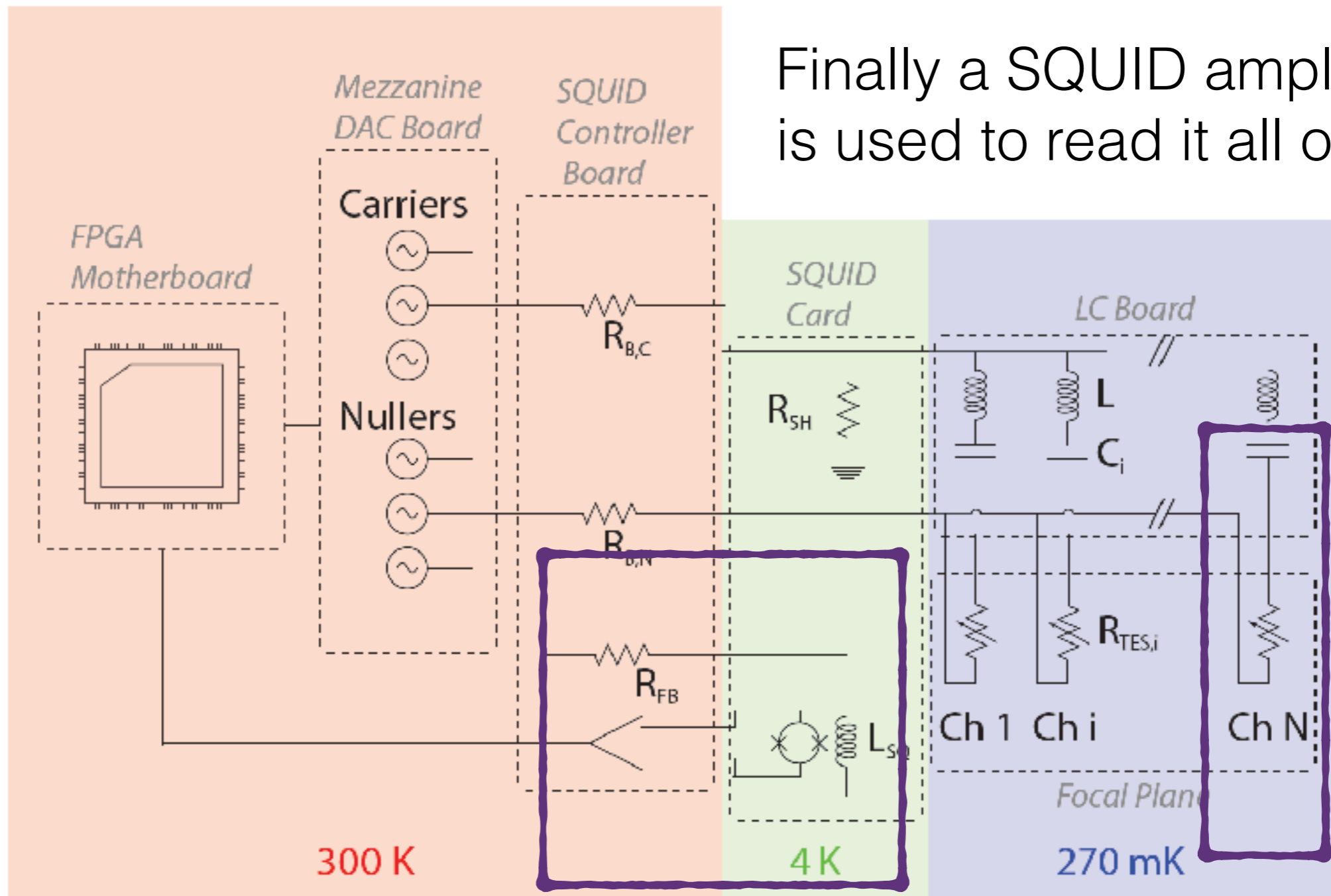


## Seeing the light: SPT



# Seeing the light: SPT

**S**uperconducting **Q**uantum **I**nterference  
**D**evice: sensitive magnetometer for  
measuring tiny (fT) magnetic fields



Finally a SQUID amplifier is used to read it all out

## Seeing the light: S4 (potential)

Multiplexing factor is good, but not great: SPT-3G has 15,234 detectors at  $68\times$  multiplexing

$0.5E6$  detectors for CMB S4, thus naively multiplexing factor should scale up by  $> 2E3$

Of-course, we are likely to split the  $0.5E6$  detectors across a few focal planes, and the readout can be further segmented

These can help us by  $O(10-100)$ , but pushing by another decade is going to very very challenging.

While 3G operations will be extremely illuminating, we recognize it is not smooth sailing

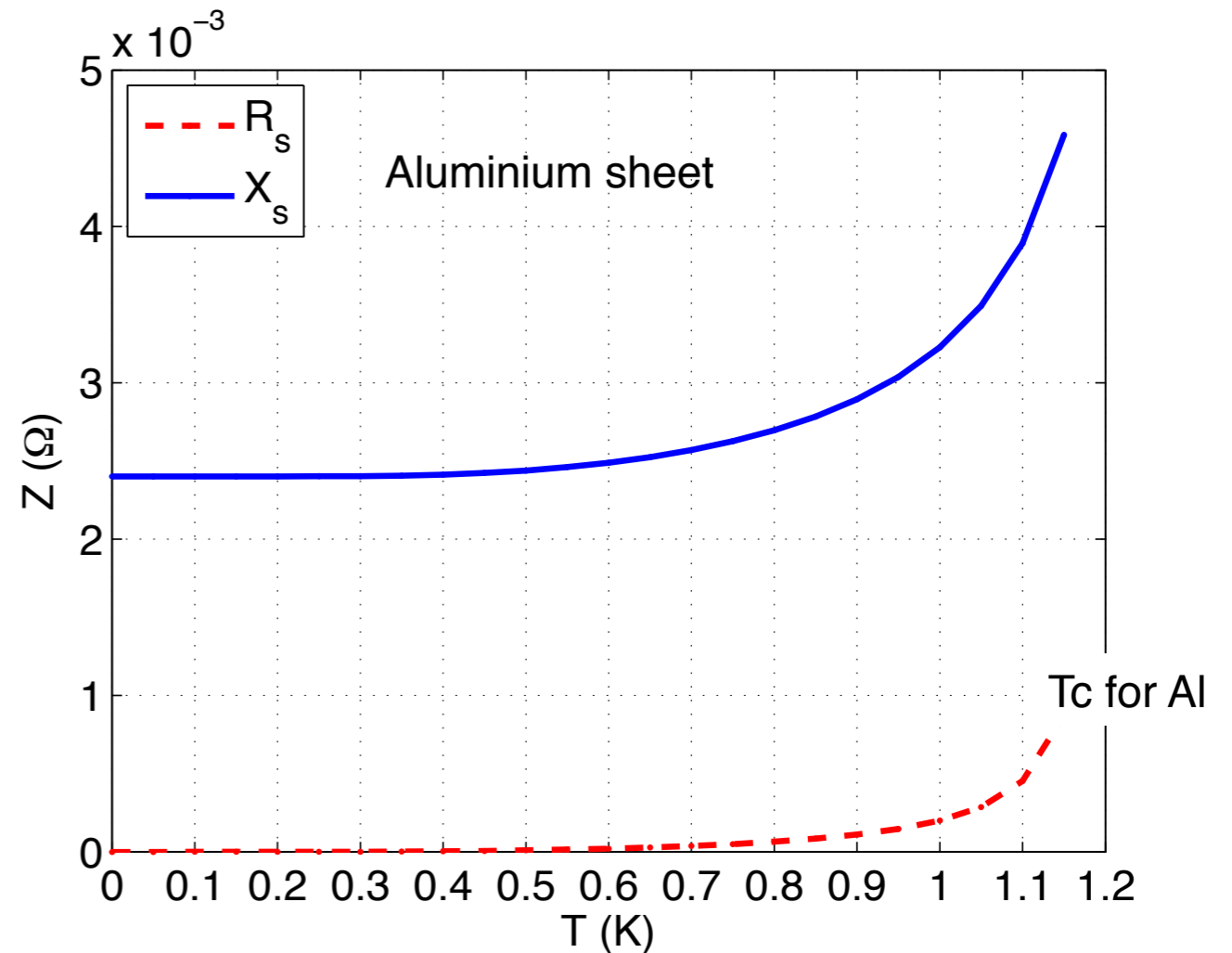
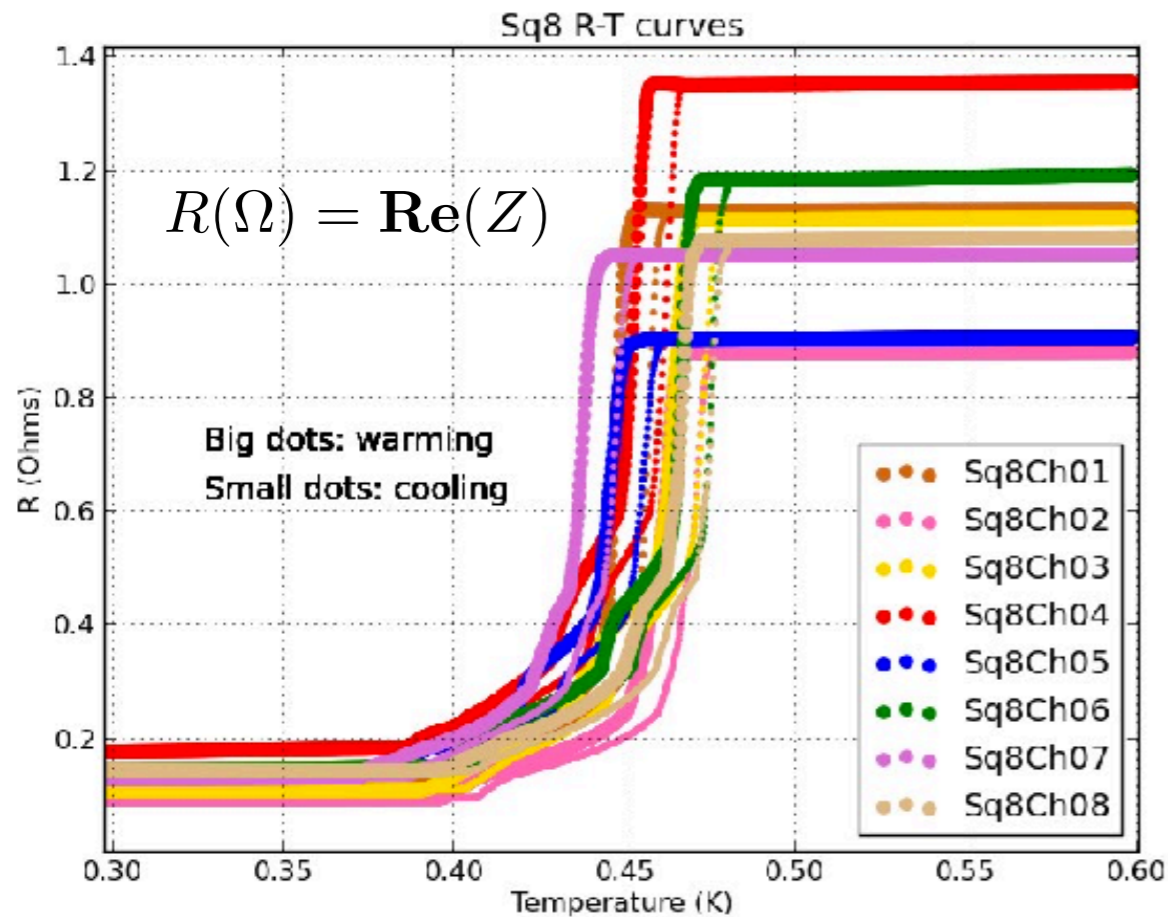
*Native multiplexing ...*



## Seeing the light: S4 (potential)

Kinetic Inductance: Superconductors electronics in all generality

$$Z(\omega) = R + i(X_c + X_L) = R + i(1/\omega C + \omega L)$$



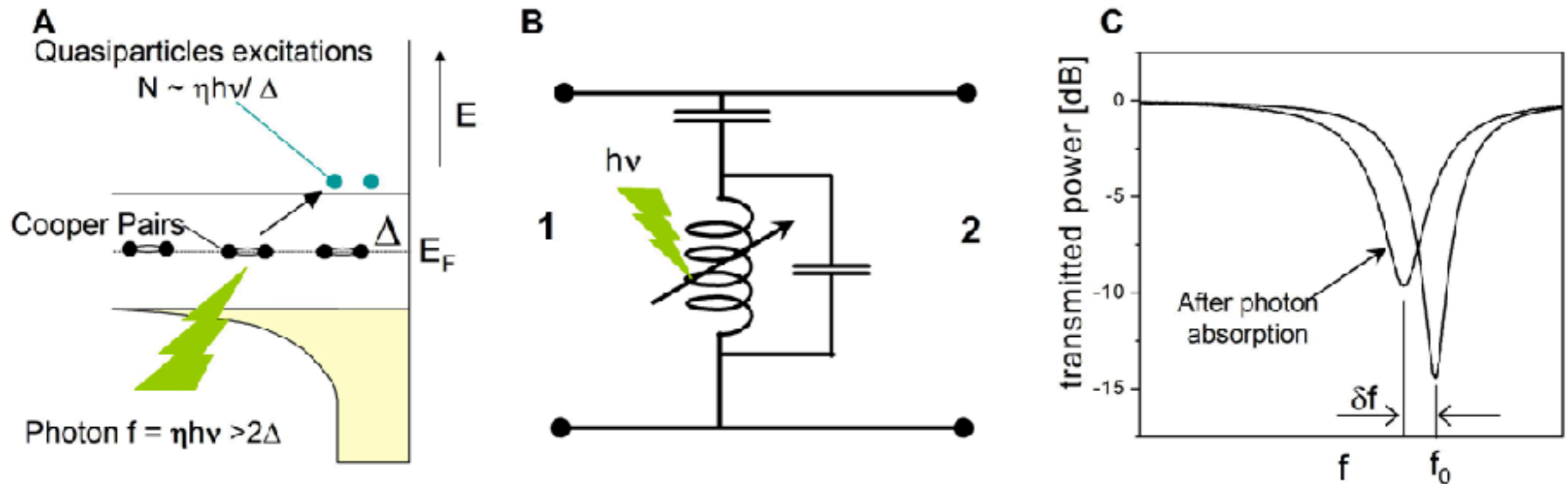
High frequency oscillating fields will see a “mass” for the Cooper

pairs, leading to a phase lag, or  $\frac{\text{Im}}{\text{Re}}(Z) > 0$

# Seeing the light: S4 (potential)

## Microwave Kinetic Inductance Detectors (MKIDs)

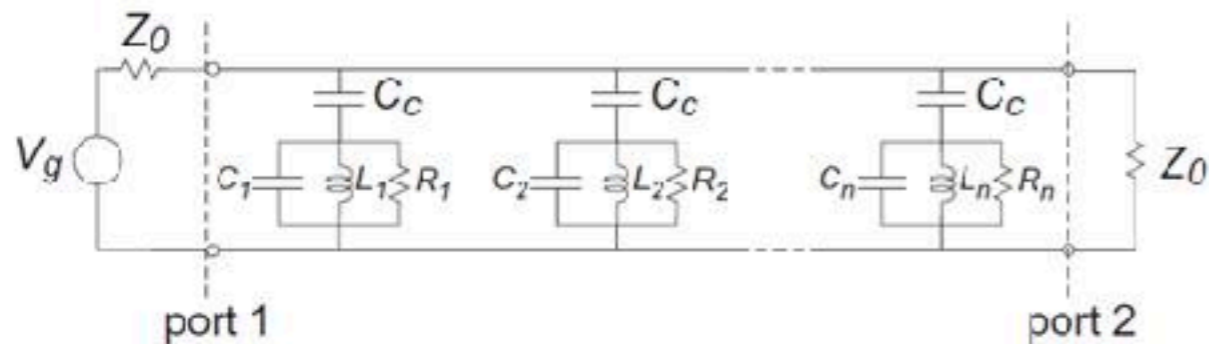
$$S_{21} = \frac{1}{1 + jQ(1 - \omega_0^2/\omega^2)}$$



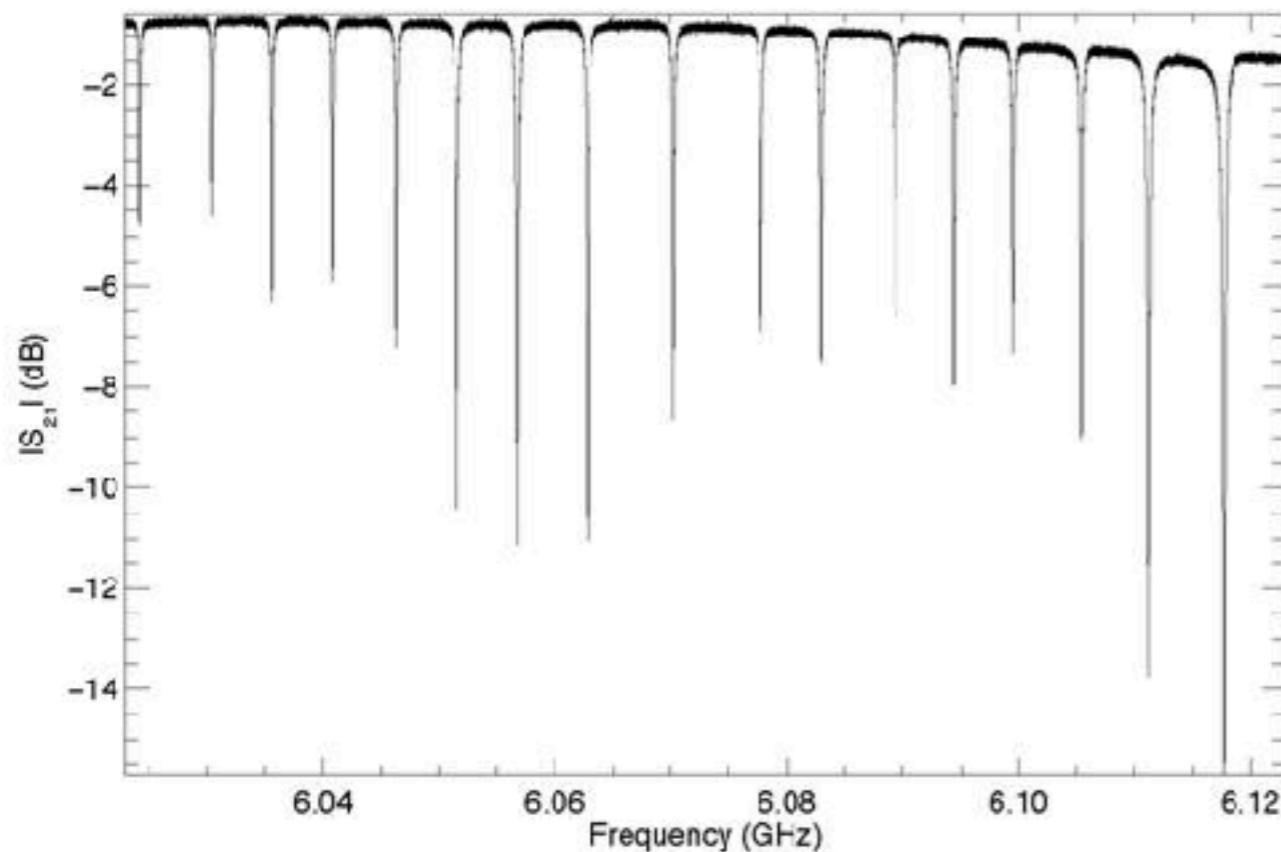
Photon absorption -> Cooper pair breaking -> Change in R & L  
-> Resonance frequency shift, phase shift, peak broadening

# Seeing the light: S4 (potential)

## Microwave Kinetic Inductance Detectors (MKIDs)



Because each detector is a mode, we can connect them in series, i.e. a comb of modes



The amplitude of a mode  $\sim$  photon power on that detector

Thus MKIDs are naturally multiplexed photon detectors

# Conclusions

We discussed how colors and patterns of the CMB can be measured

Rest of 448 course has shown the physics behind these fluctuations

General sensitivity calculations were presented

Types of technologies were outlined

*We did not cover: foreground subtraction, map making algorithms and parameter estimations*

*Questions ?*